

# *Radio Research Directions*

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# Outline

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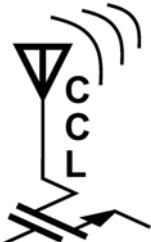
- **Introduction**
- **Millimeter-Wave Transceivers**
  - Applications
  - Challenges
  - Examples
- **Cognitive Radios**
  - Challenges
- **Conclusion**



# Nature of Our Radio Research

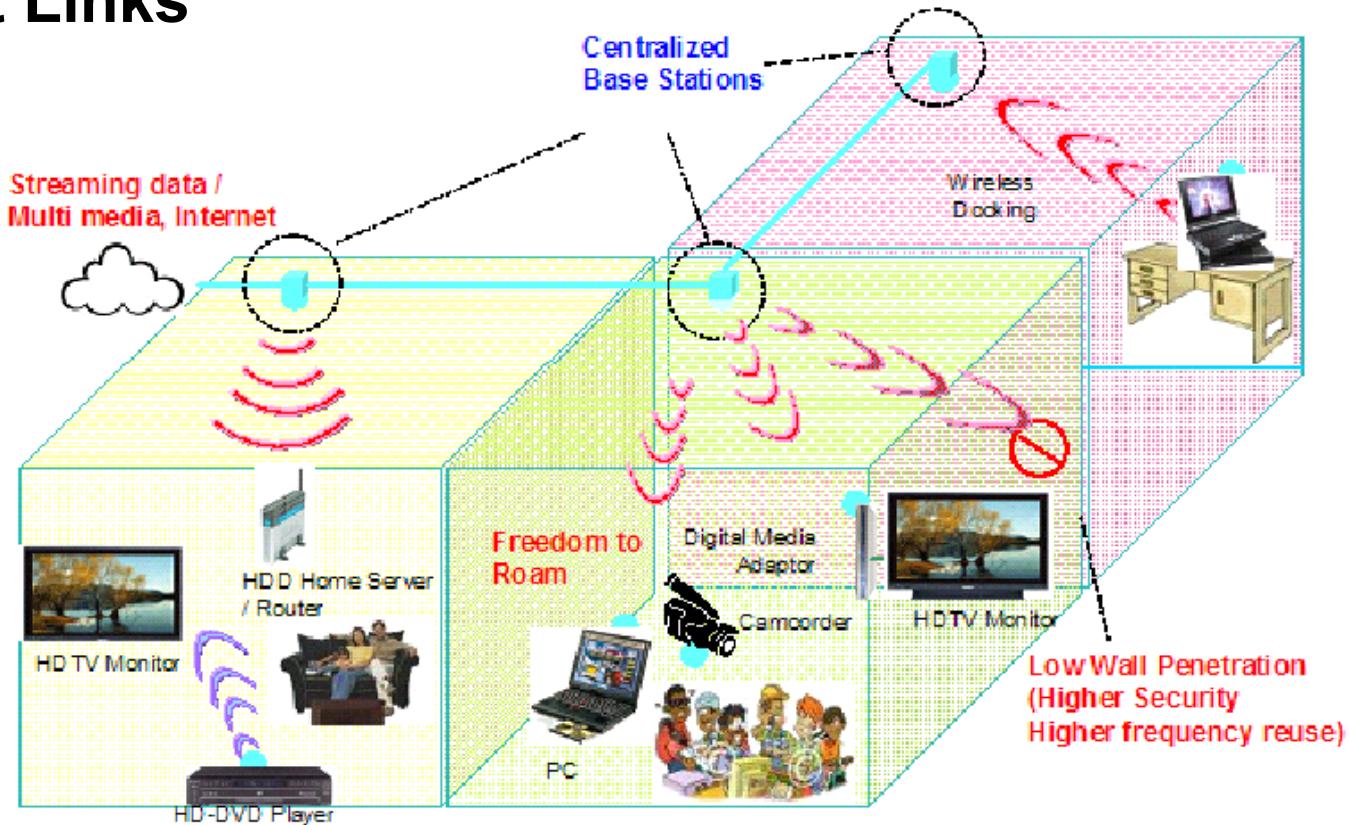
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- Target highly-integrated transceivers with minimal number of external components
- Address tough problems: higher frequency, wider bandwidth, lower power consumption, ...
- Develop new architectures, circuits, devices that solve these problems
- Realize ideas in standard CMOS technology and verify by experimentation
- Examples of past work:
  - 900-MHz/1.8-GHz Transceivers for Cellular Telephony
  - 2.4-GHz and 5.2-GHz WLAN Transceivers
  - 3-6 GHz UWB Transceivers
  - 5-GHz RX for MIMO
  - 60-GHz Transceivers



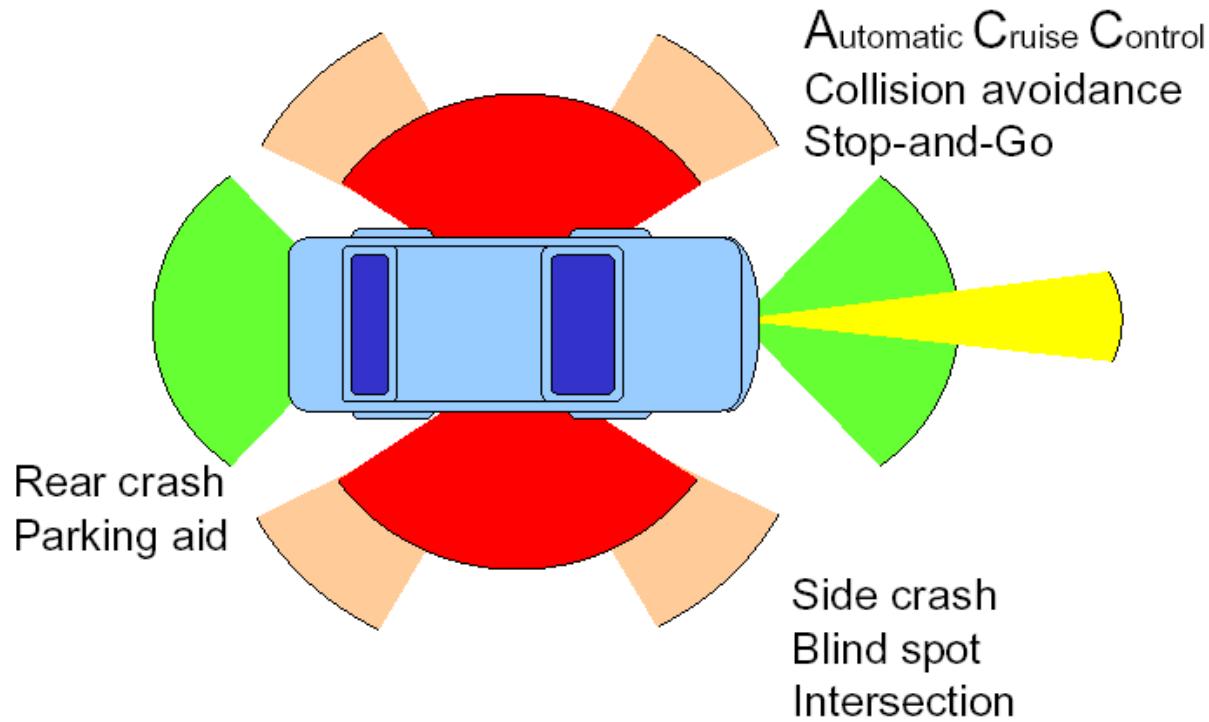
# Why the interest in mmWaves?

- Unlicensed band: 57 GHz – 64 GHz offers possibility of high-data rate communications:
  - High-Definition Video Streaming
  - Fast Links



# Why the interest in mmWaves?

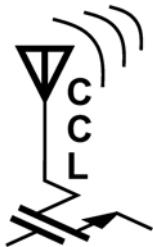
- Automotive Radar (60-77 GHz)



# Why the interest in mmWaves?

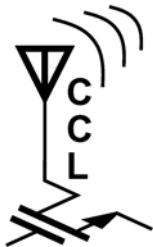
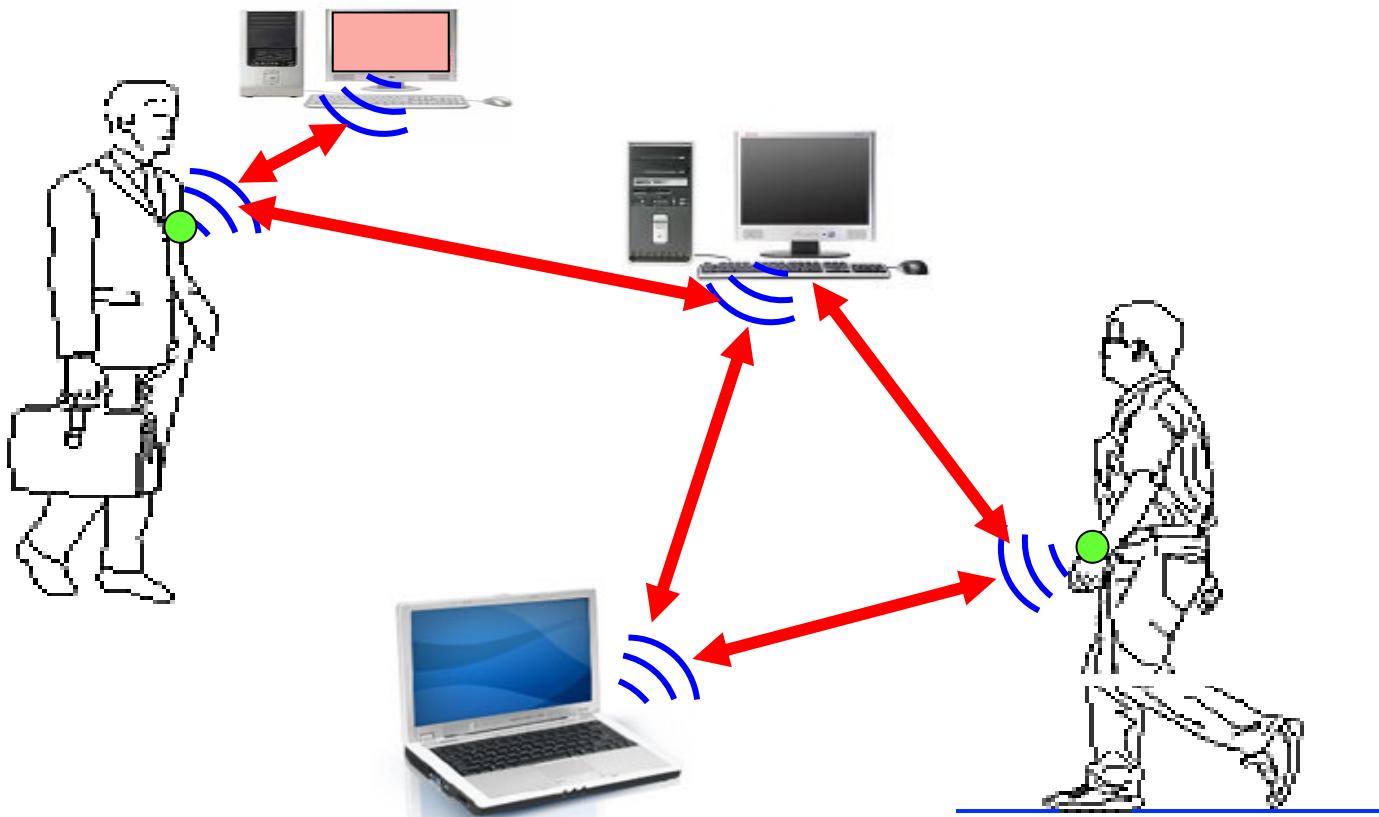
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- mmWave Imaging (> 100GHz)

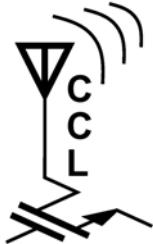
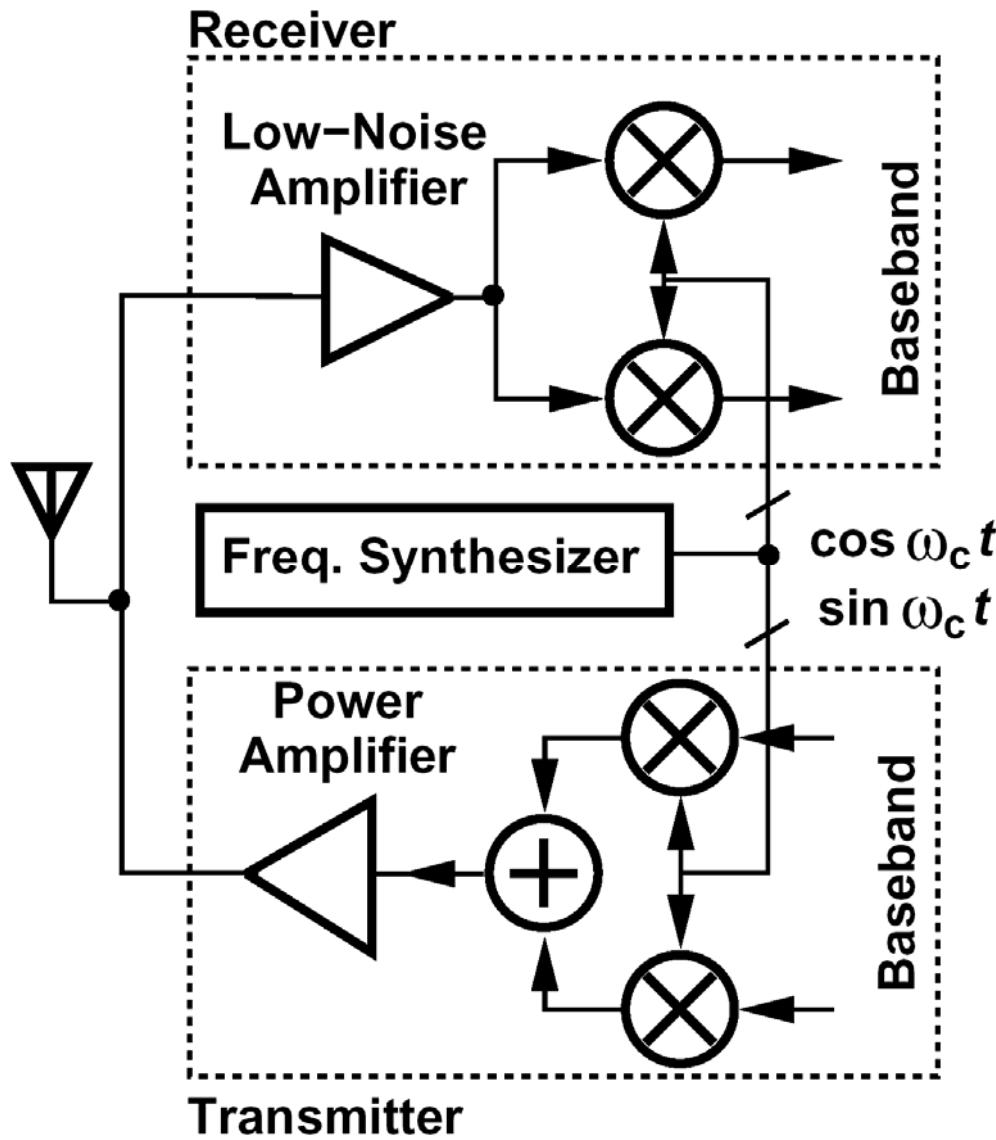


# Networks with High Redundancy

- Line-of-sight propagation a serious issue

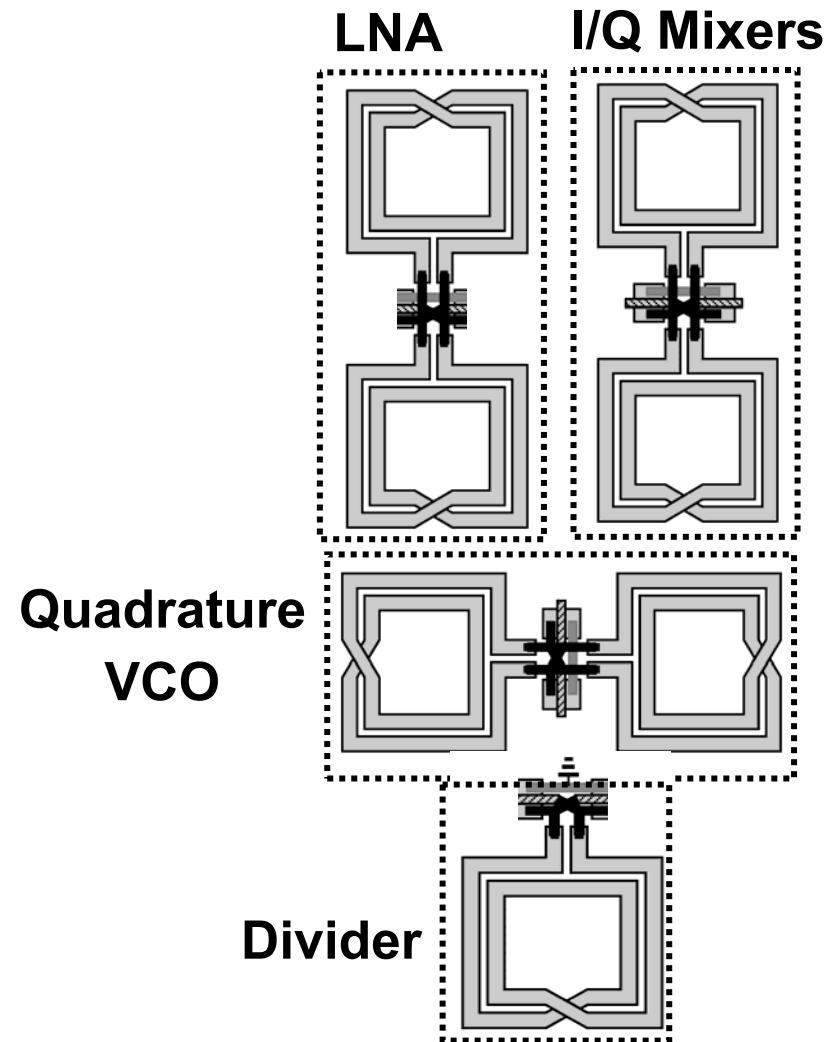
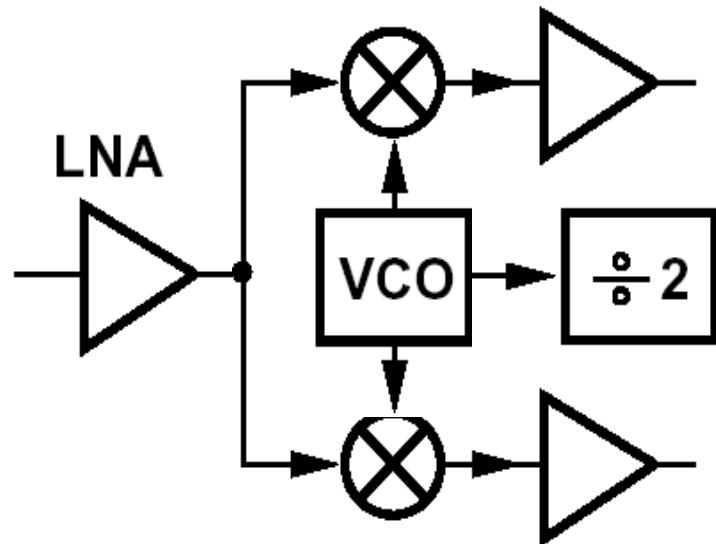


# A Few Words for the RF-Challenged



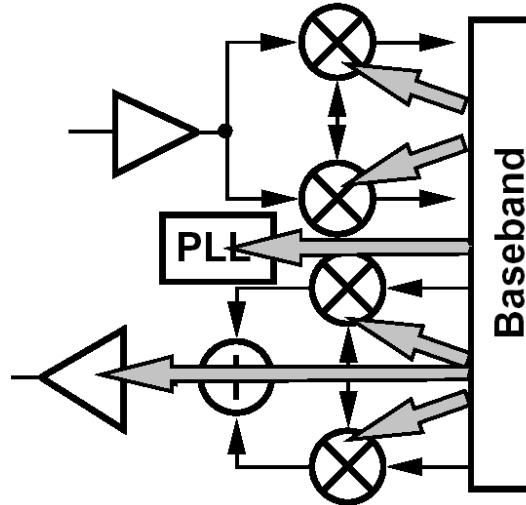
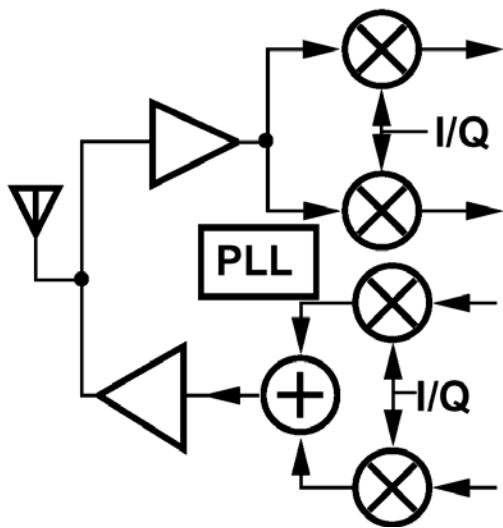
# Architecture-Level Challenges

- LO (I/Q) Generation
- LO Division
- LO Distribution



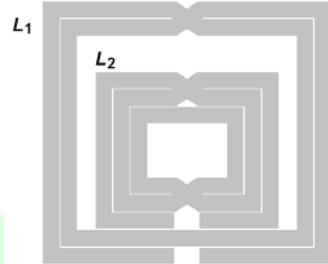
# Innovation at All Levels

## System

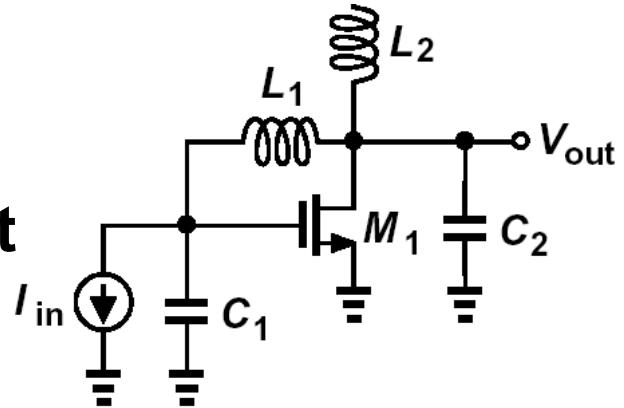


## Architecture

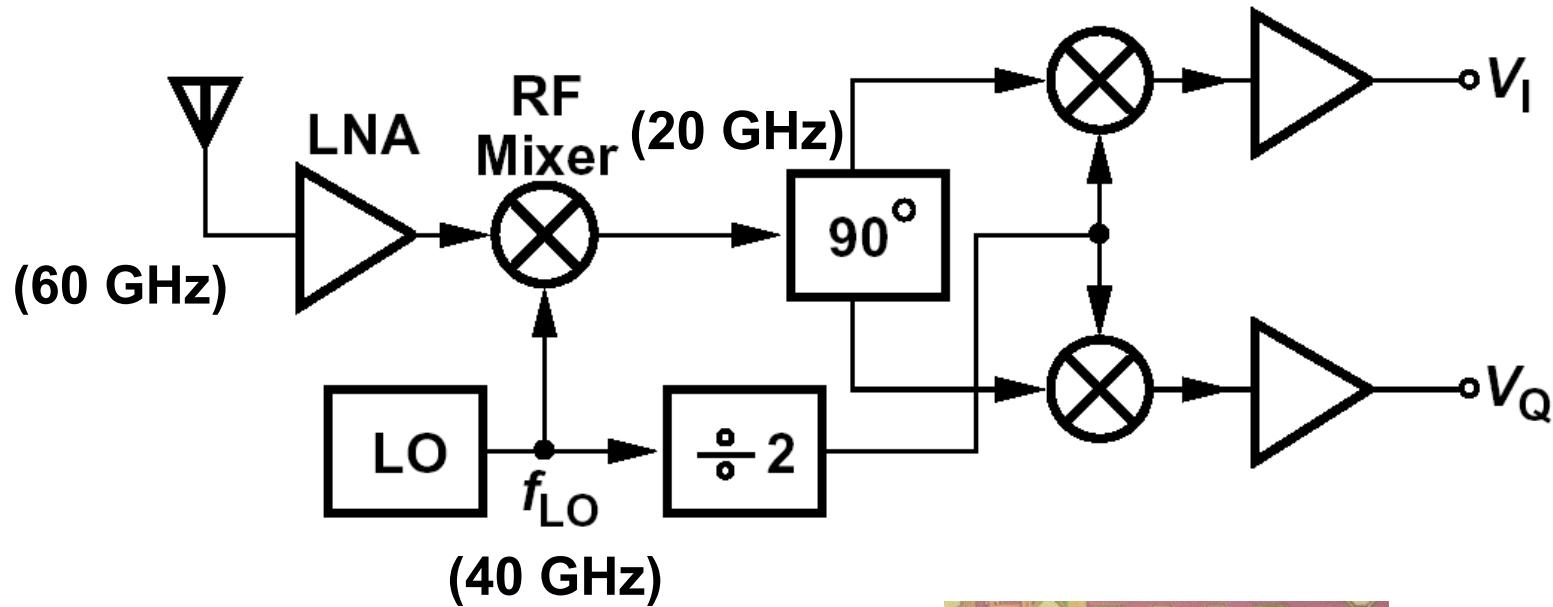
## Device



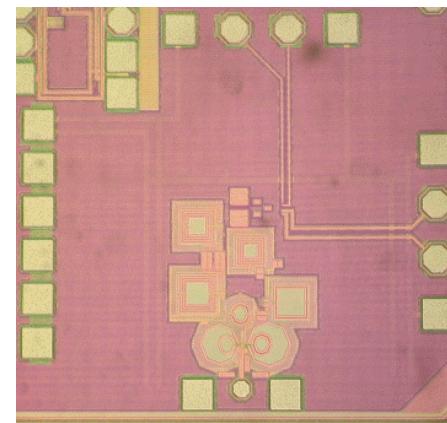
## Circuit



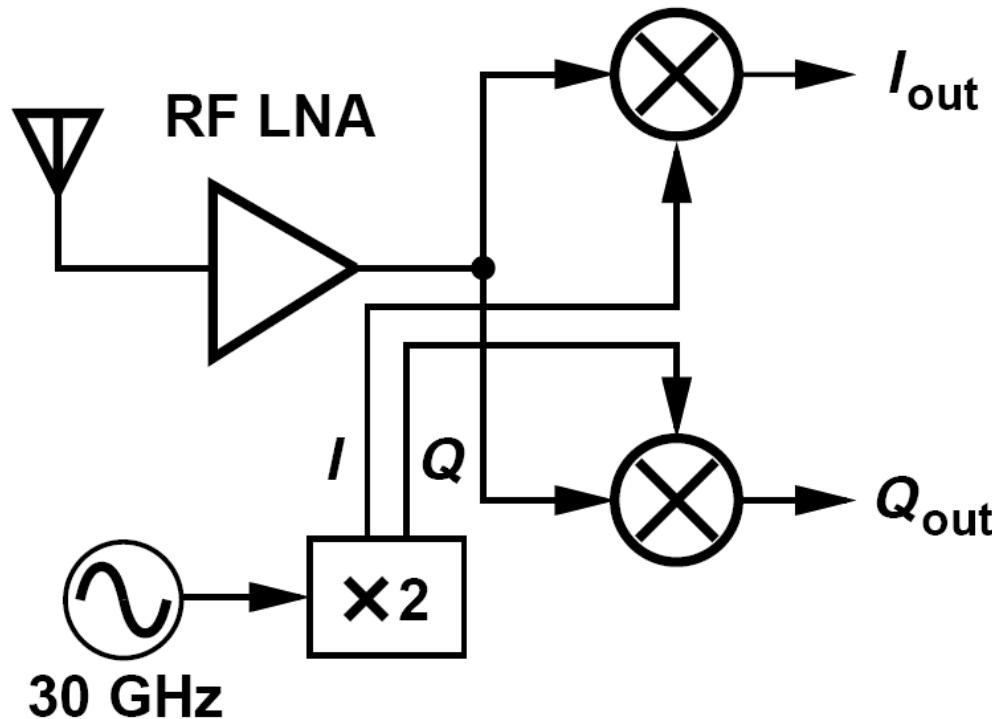
# Our Second-Generation 60-GHz RX



[B. Razavi, ISSCC 07]



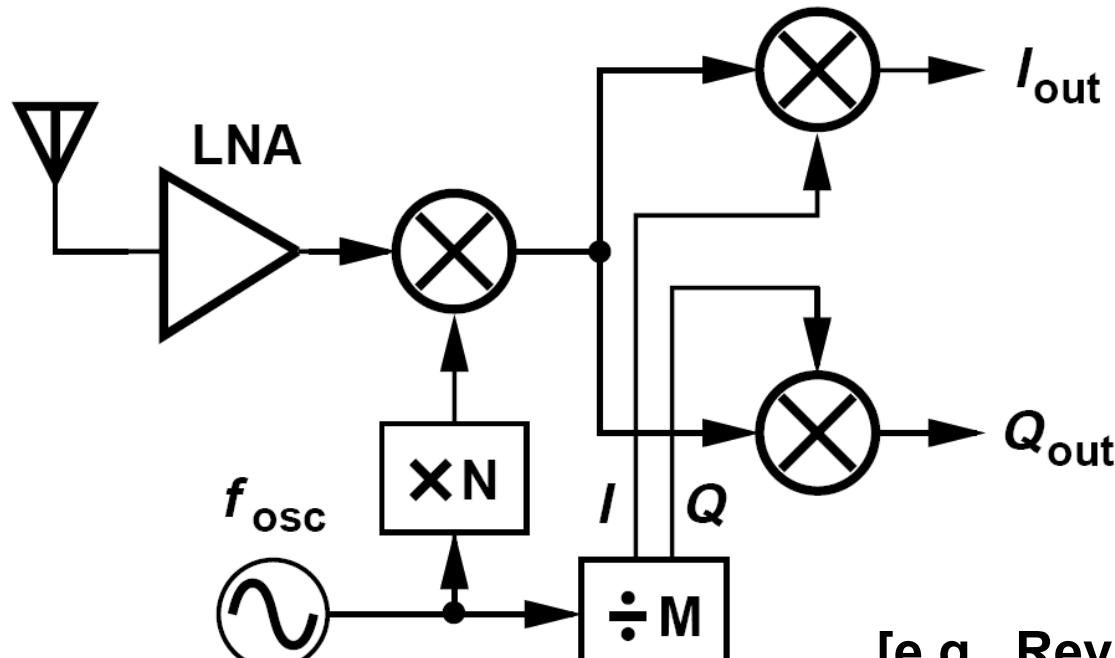
# Direct-Conversion RX with 30-GHz LO?



- Quadrature generation is difficult.
- Distribution is difficult.
- Need “synthesizer-friendly” transceivers.



# Heterodyne Receiver

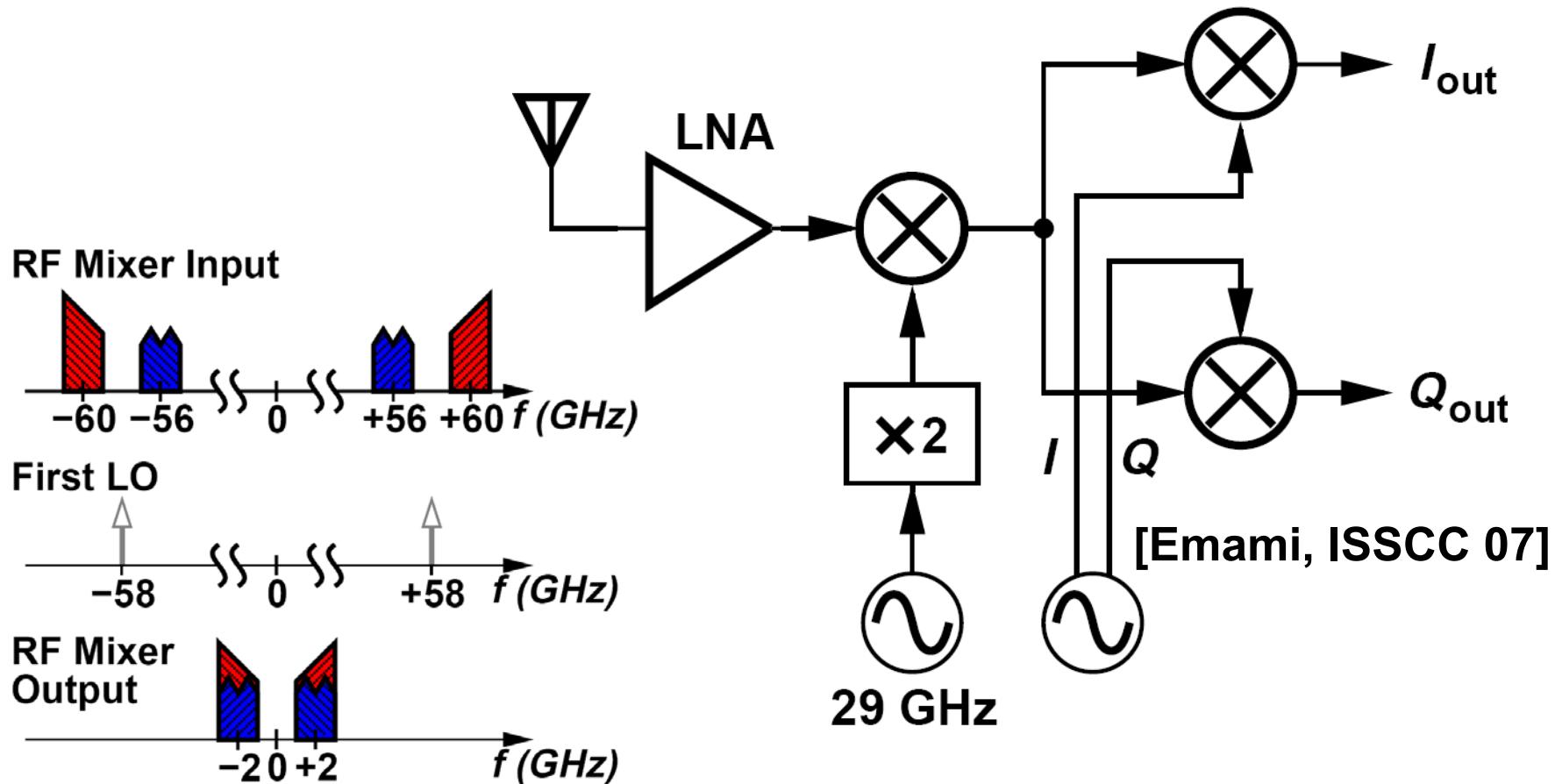


[e.g., Reynolds,  
JSSC, Dec 06]

- Multiplier has high loss and needs its own inductor.



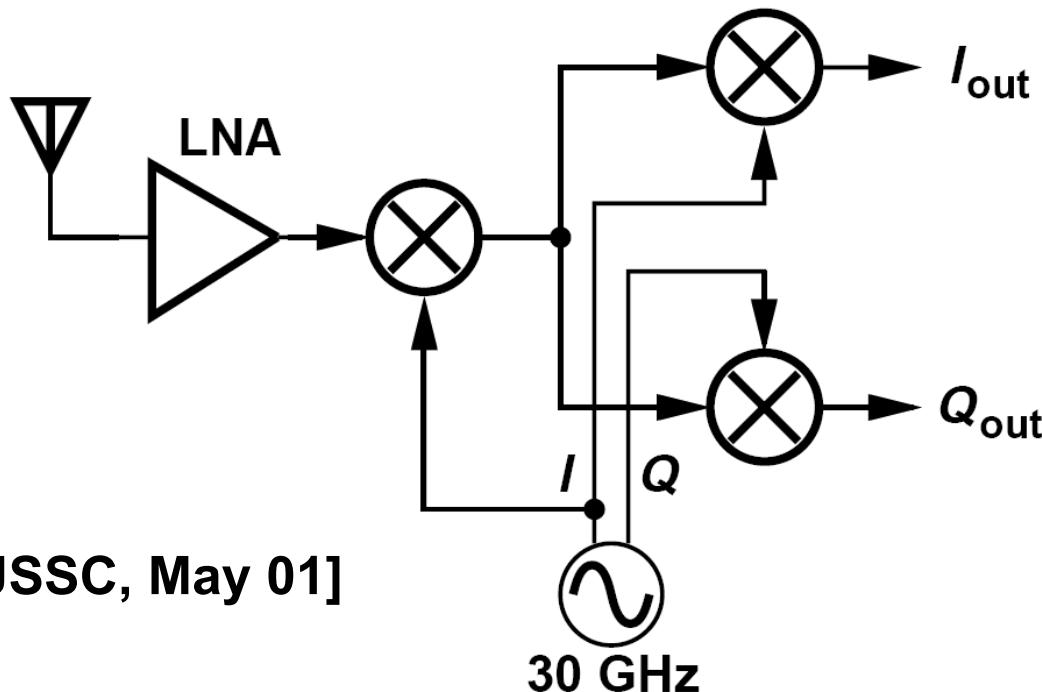
# Problem of Low-IF Heterodyne



- Image of the first mixer is in the band.
- Receiver NF is increased by  $\sim 3$  dB.



# Example of Synthesizer-Friendly Receiver



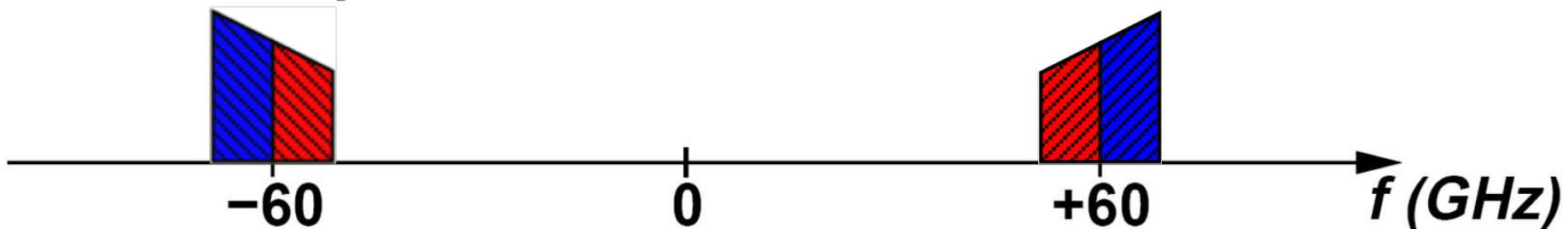
[Razavi, JSSC, May 01]

- No extra divider/multiplier needed.
- Image is at DC.
- But,
  - Third harmonic of LO causes corruption.
  - LO-IF feedthrough may desensitize the IF mixers.
  - 1/f noise is upconverted in RF mixer.

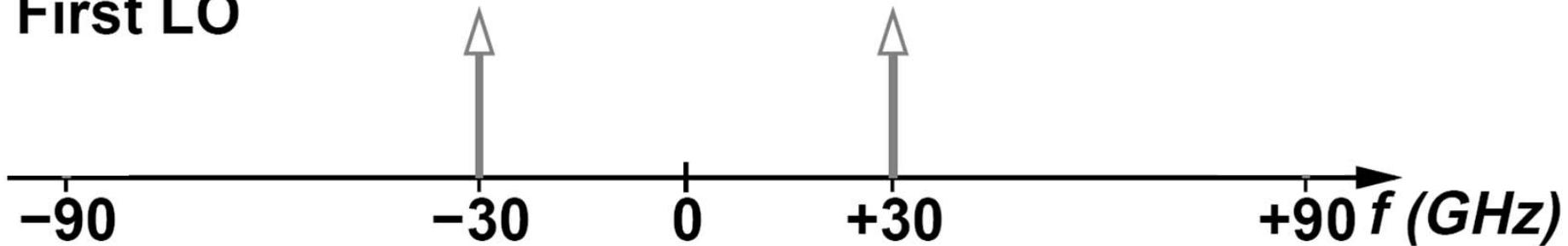


# Problem of LO Third Harmonic

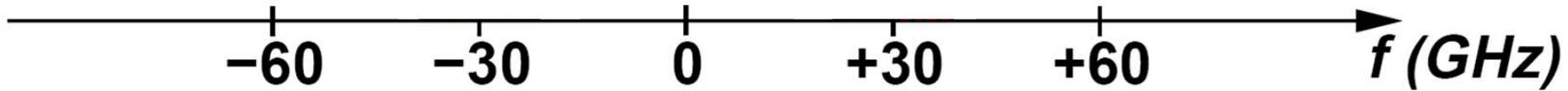
RF Mixer Input



First LO



RF Mixer  
Output



# Analysis

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$$x_{RF}(t) = \Re \left[ x_{BB}(t) e^{+j2\pi f_{RF}t} \right]$$

$$LO = \cos \omega_{LO} t + \alpha \cos 3\omega_{LO} t \quad \alpha \approx 1/3$$

$$x_{IF}(t) = \Re \left[ \frac{x_{BB}(t) + \alpha x_{BB}^*(t)}{2} e^{+j2\pi f_{IF}t} \right]$$

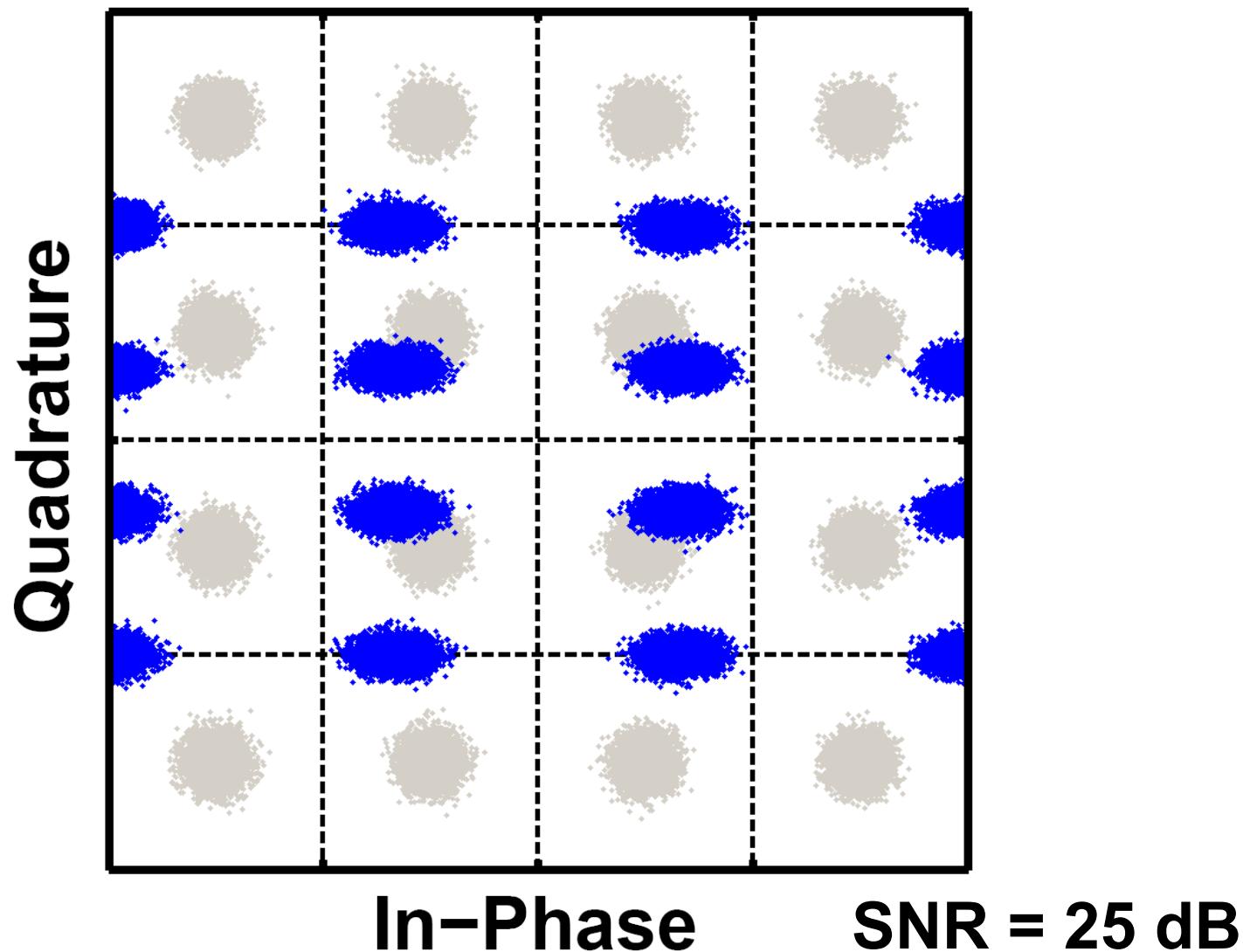
$$x_{out}(t) = k \left[ x_{BB}(t) + \alpha x_{BB}^*(t) \right]$$

$x_{BB}(t)$  : **wanted signal**       $\mathcal{F}\{x^*(t)\} = X^*(-f)$

$x_{BB}^*(t)$ : **mirrored replica of the signal**

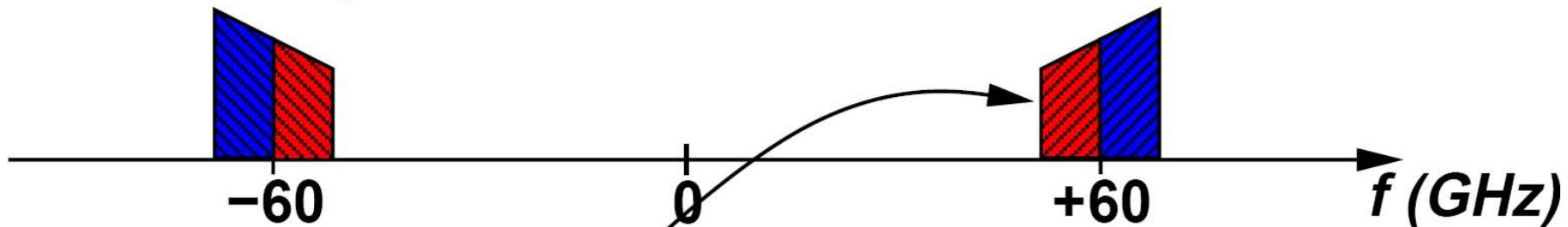


# 16-QAM Constellation

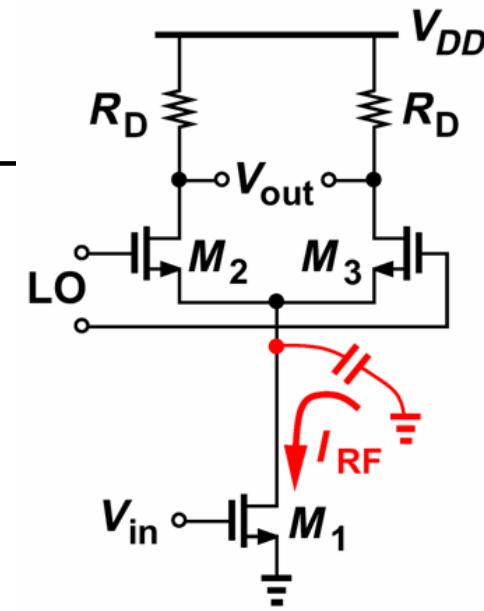
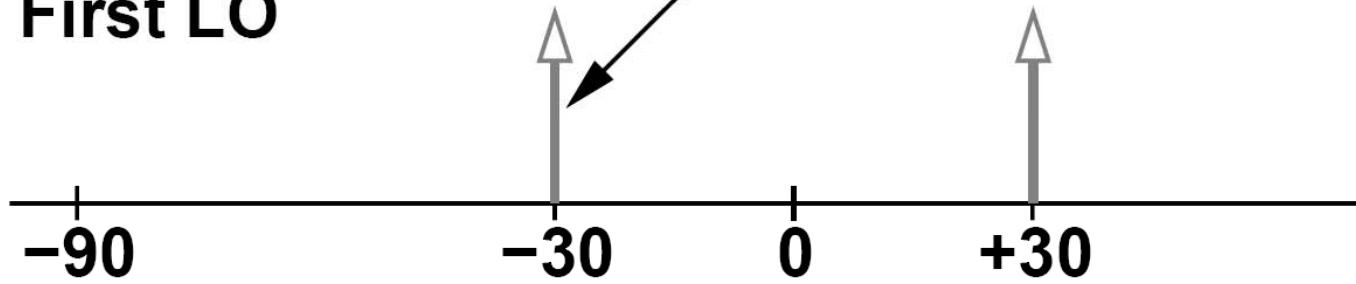


# Linearize LO Port?

## RF Mixer Input

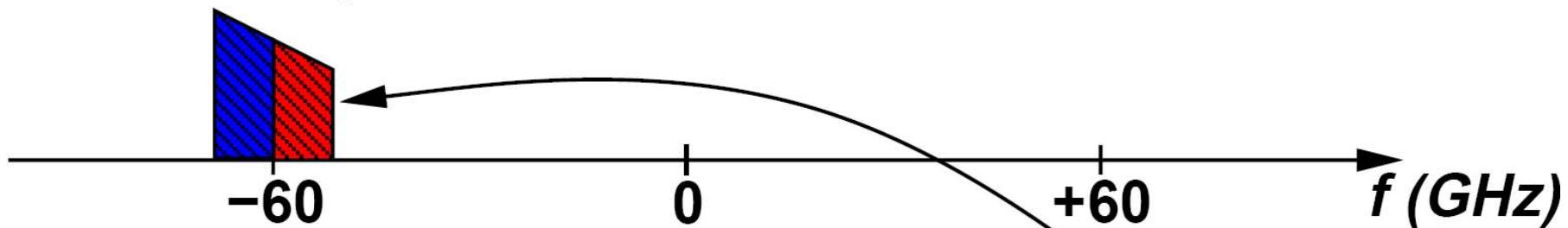


## First LO

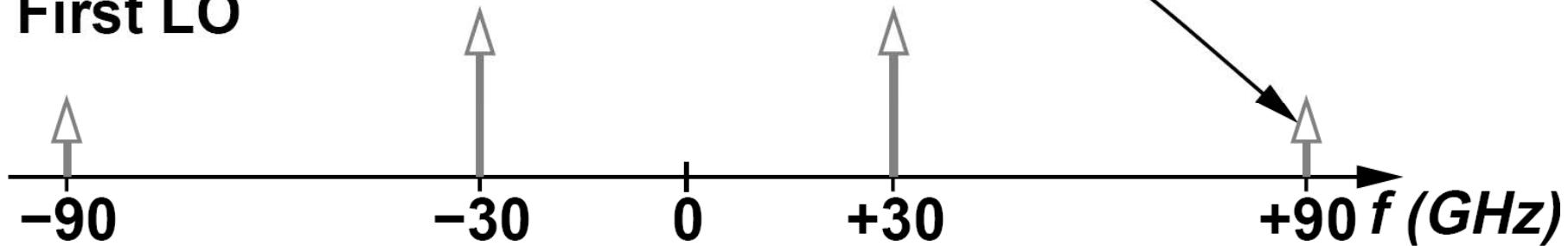


# Alternative Solution

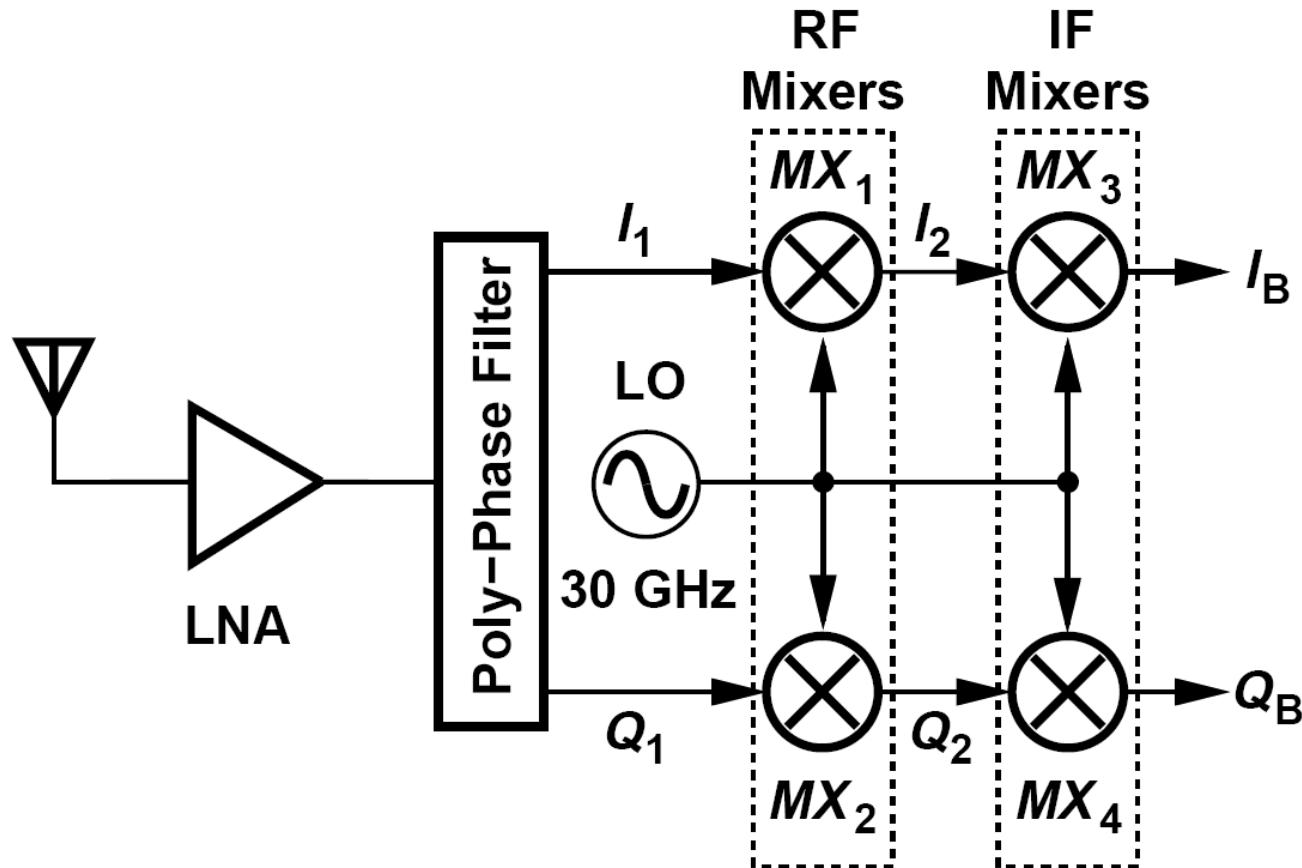
## RF Mixer Input



## First LO



# Proposed Receiver Architecture



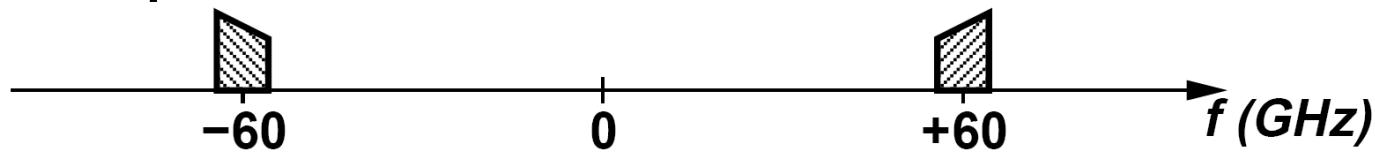
- Lowest possible LO frequency (without multiplication).
- No quadrature LO phases required.



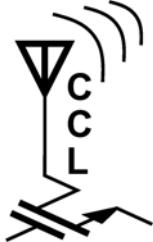
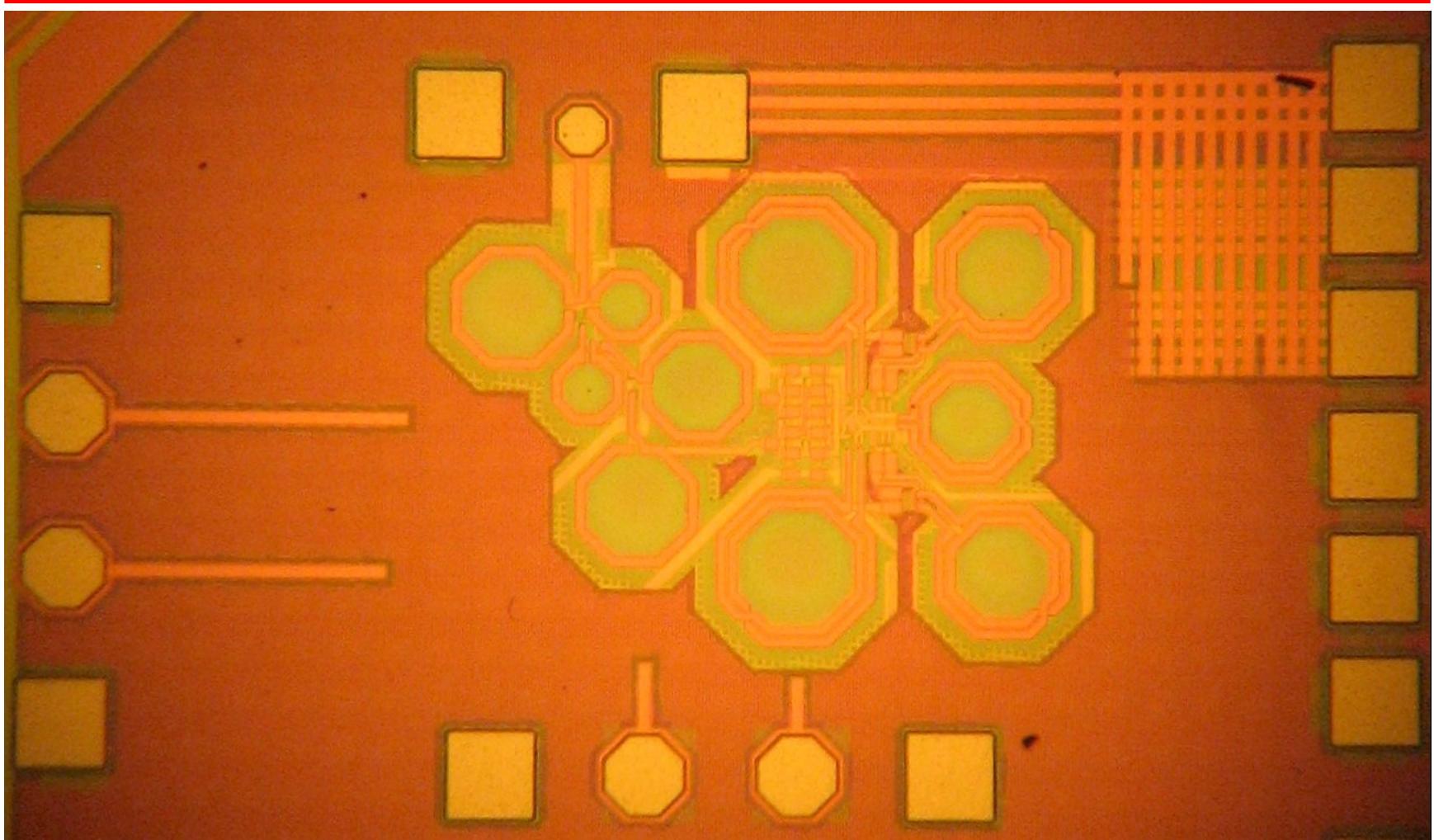
# Receiver Spectra



RF input



# Die Photograph

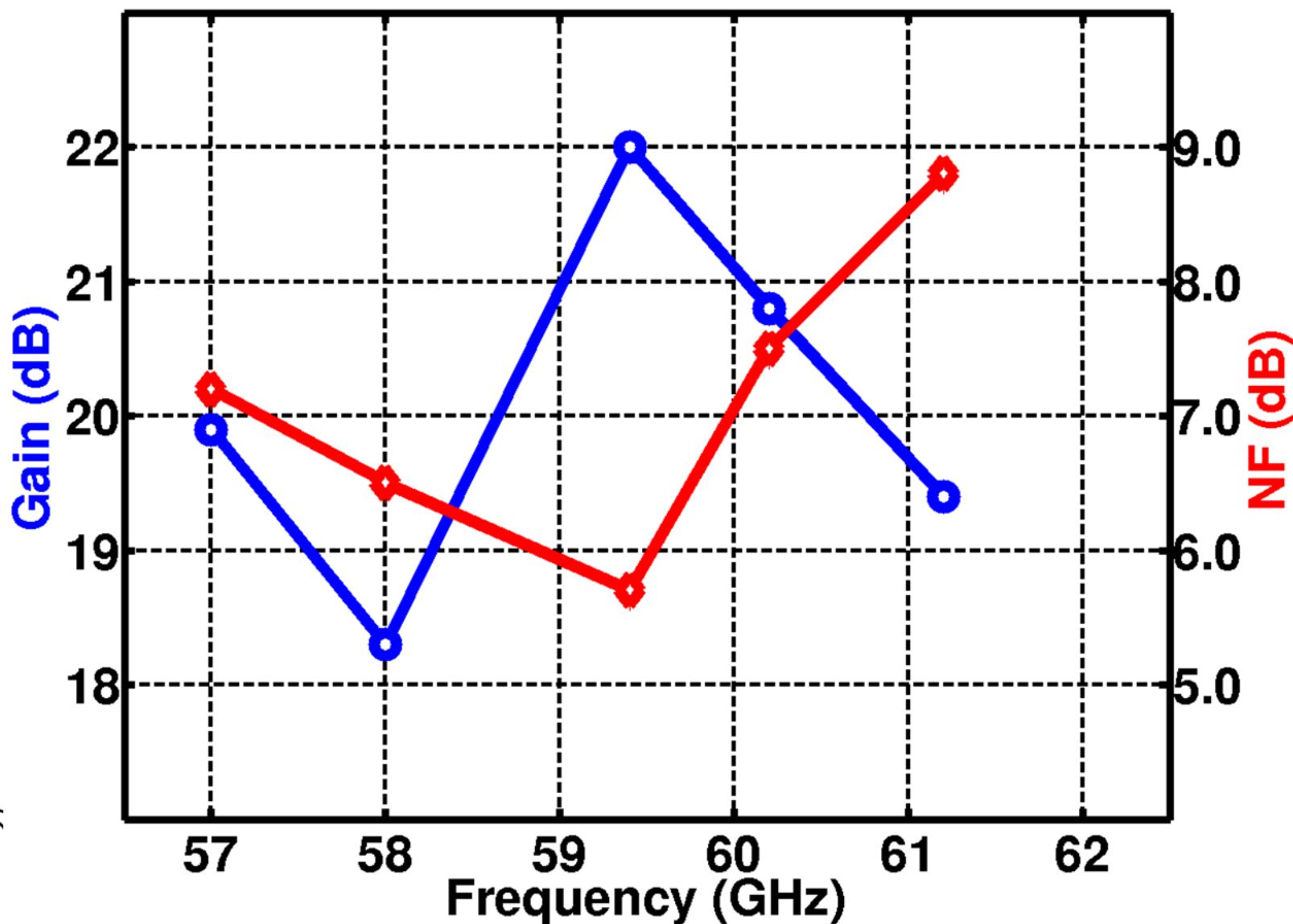


**Fabricated in TSMC's 90-nm CMOS technology.**

**Active area: 500  $\mu\text{m}$  x 370  $\mu\text{m}$**

***Communication Circuits Laboratory***

# Measured NF and Gain



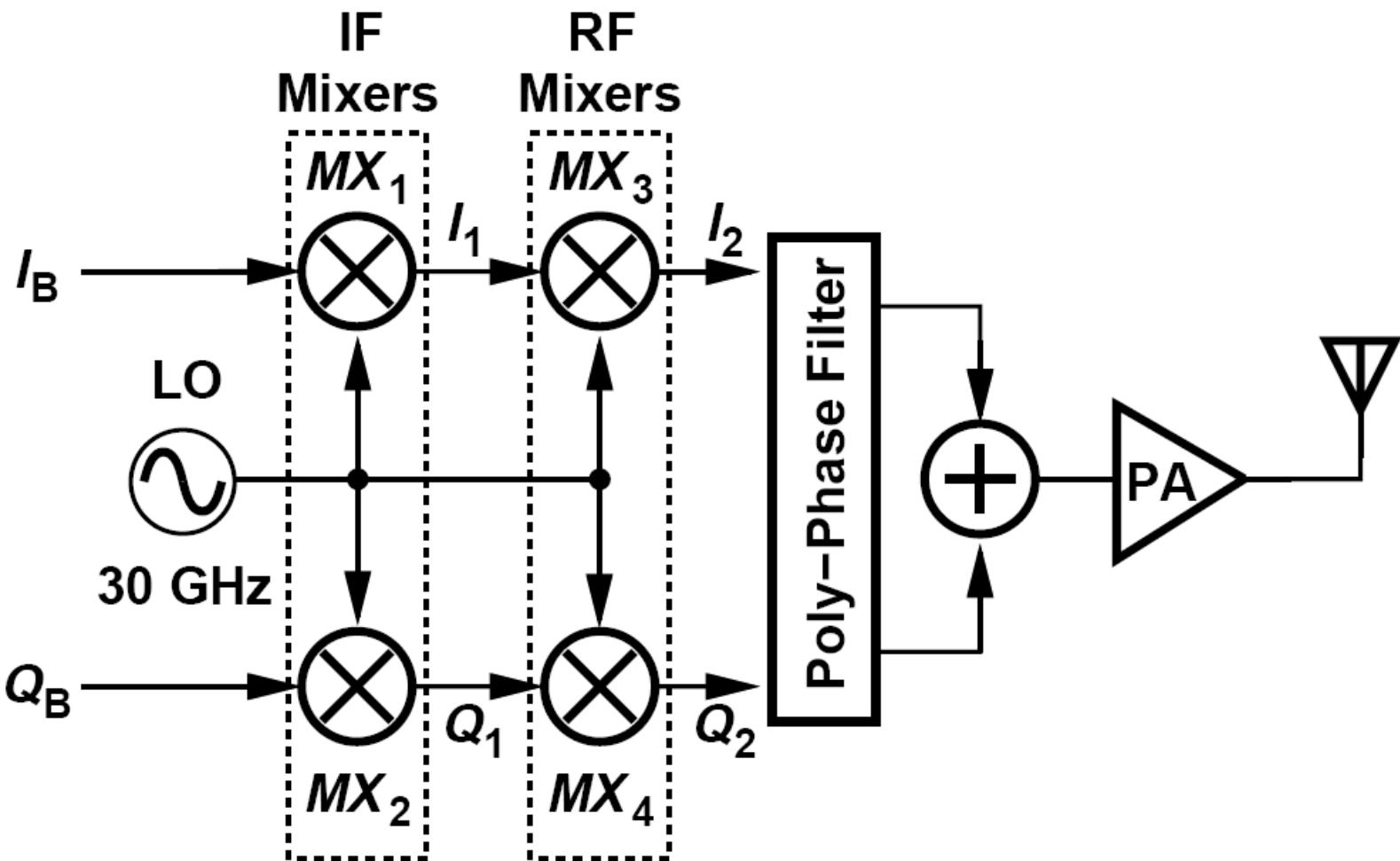
# Comparison

	Receiver in [3]	Receiver in [1]	This work
<b>Noise Figure (dB)</b>	<b>10.4-11</b>	<b>6.9-8.3</b>	<b>5.7-8.8</b>
<b>Gain (dB)</b>	<b>9.5-12</b>	<b>26-31.5</b>	<b>18.3-22</b>
<b>P<sub>1dB</sub> (dBm)</b>	<b>-15.8</b>	<b>-25.5</b>	<b>-27.5</b>
<b>LO Leakage to Input (dBm)</b>	<b>N/A</b>	<b>-47</b>	<b>-65</b>
<b>I/Q Mismatch</b>	<b>N/A</b>	<b>6.5°/ 1.5dB</b>	<b>2.1°/ 1.1dB</b>
<b>LO Phase Noise (dBc/Hz @ 1-MHz offset)</b>	<b>-86</b>	<b>-95</b>	<b>-87</b>
<b>Power Dissipation (mW)</b>	<b>77</b>	<b>80</b>	<b>36</b>
<b>LNA</b>			<b>9</b>
<b>Mixers</b>			<b>23</b>
<b>Oscillator</b>			<b>4</b>
<b>Supply Voltage (V)</b>	<b>1.2</b>	<b>1.8</b>	<b>1.2</b>
<b>CMOS Technology</b>	<b>0.13-μm</b>	<b>90-nm</b>	<b>90-nm</b>

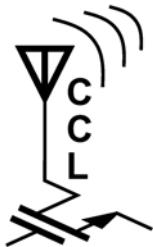


[1] B. Razavi, ISSCC '07  
[3] S. Emami et al, ISSCC '07

# Transmitter Architecture

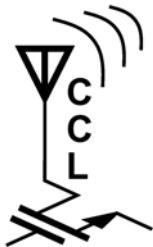
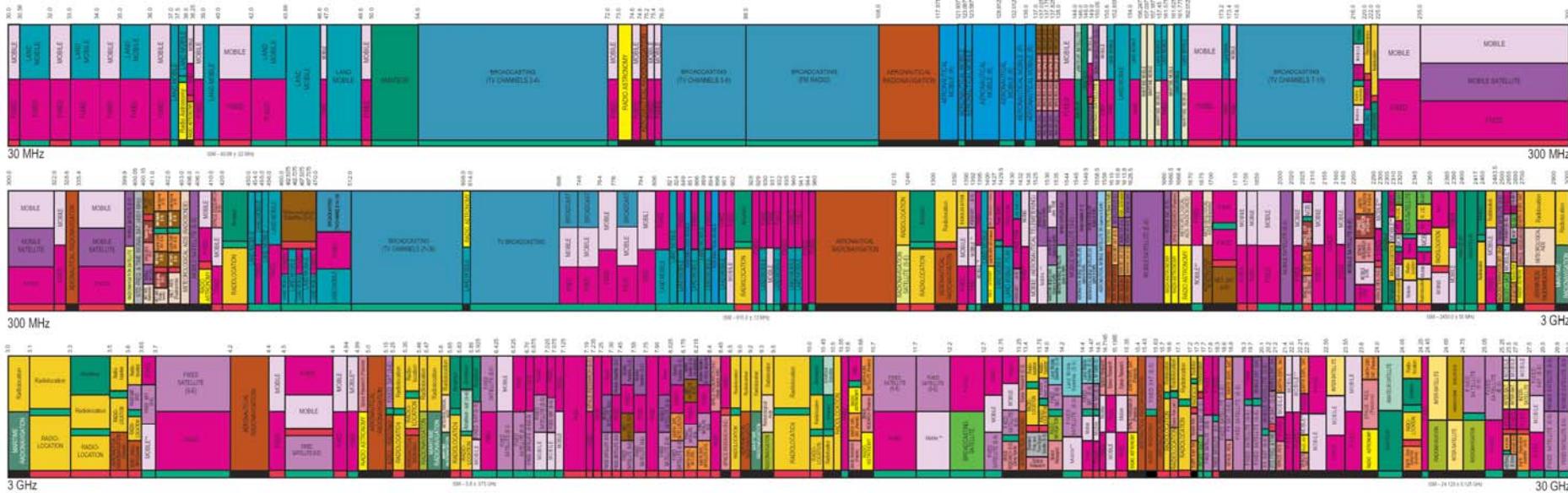


- Does not require quadrature LO



# Cognitive Radio

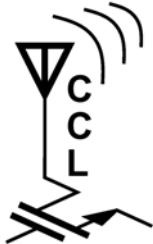
Detect and use unoccupied channels.



# RF/Analog PHY Design Issues

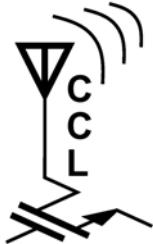
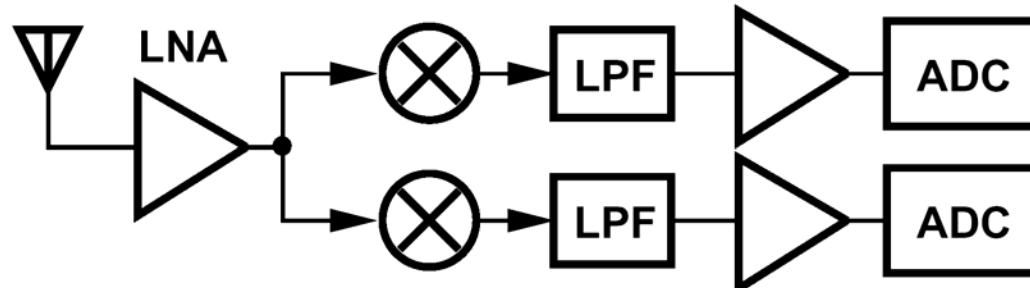
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- Spectrum Sensing
- RX Path
- TX Path
- Frequency Synthesis



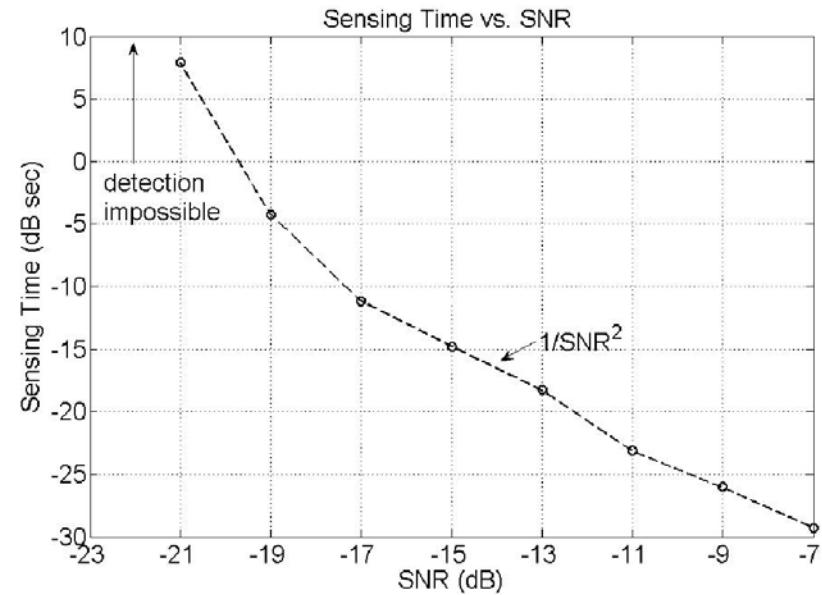
# Spectrum Sensing (I)

- Three Techniques:
  - Energy Detection
  - Pilot Detection
  - Signal Feature Detection
- Need to measure  $\text{SNR} \sim -20 \text{ dB} \rightarrow$ 
  - Accurate calibration of RX NF  
(i.e., need a tone with accurate amplitude)
  - Need enough gain to raise RX noise to well above 1 LSB of ADC



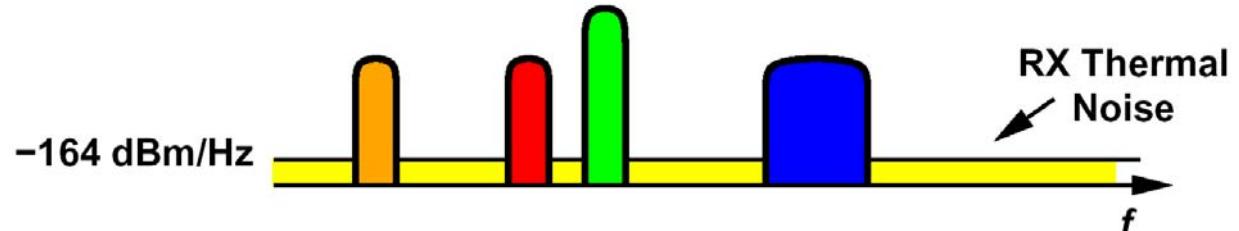
# Spectrum Sensing (II)

- Channel-by-Channel Sensing
  - Relaxed ADC design (~3 bits)
  - Takes forever.  
(e.g., 4-MHz QPSK channel:  
30 ms for SNR=-17 dB)
  - May not know the center  
or bandwidth of channel.



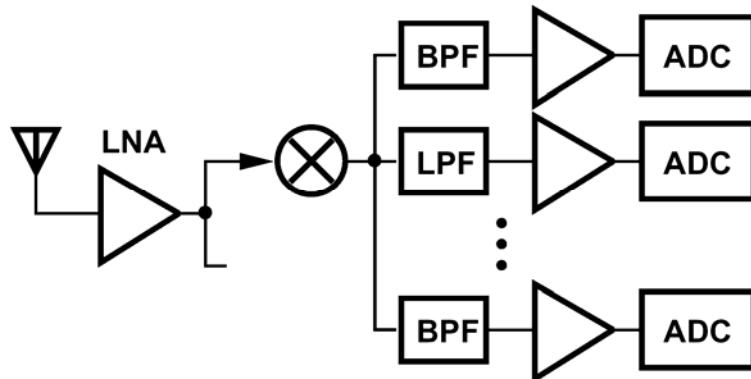
[Cabric, PhD Diss., UCB]

- Block Downconversion Sensing
  - Proportionally faster
  - But ADC BW and resolution  
much tougher



# Spectrum Sensing (III)

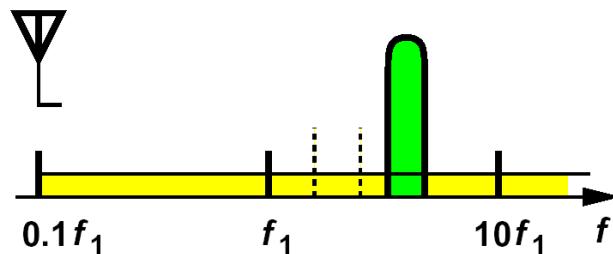
- Two-Step Sensing:
  1. ADC takes a snapshot of a block of channels and determines “potentially-unoccupied” channels.
  2. Baseband filters “zoom in” onto those channels and multiple ADCs digitize them.



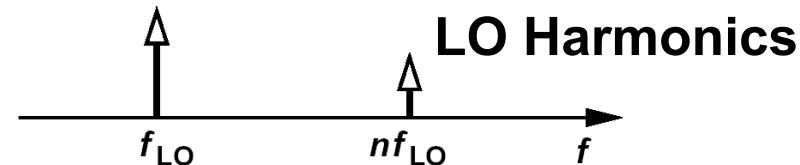
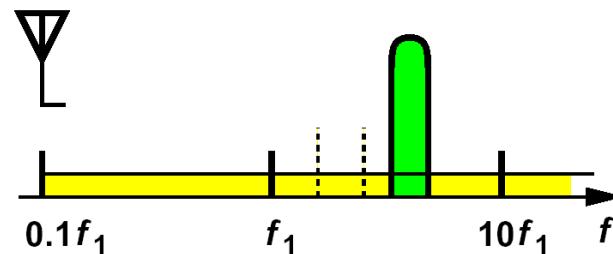
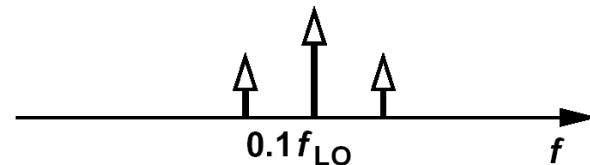
- Given certain blocker levels, what ADC resolution suffices for the first step?
- What criteria should be used to determine “potentially-unoccupied” channels?



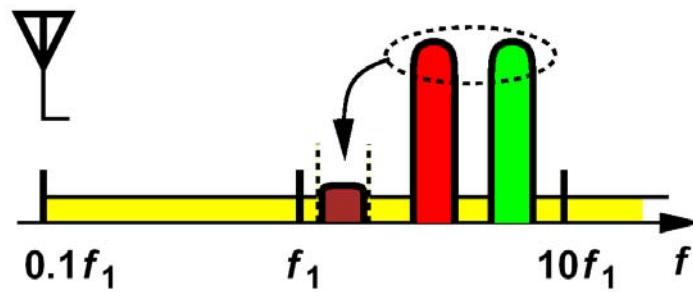
# Effect of Spurs, Harmonics, and Other Blemishes



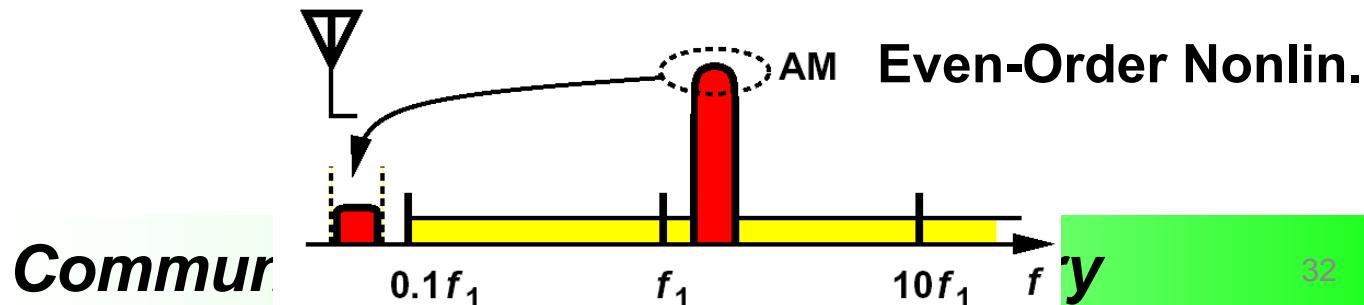
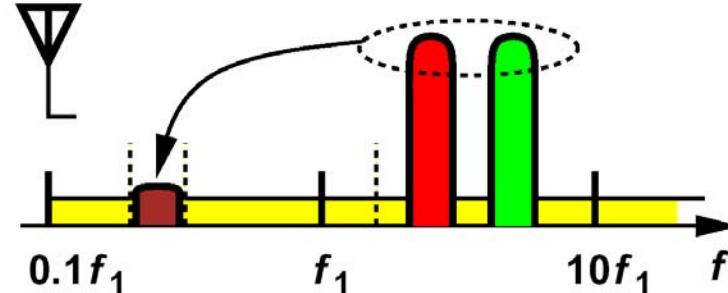
LO Spurs



Odd-Order Nonlin.



Even-Order Nonlin.



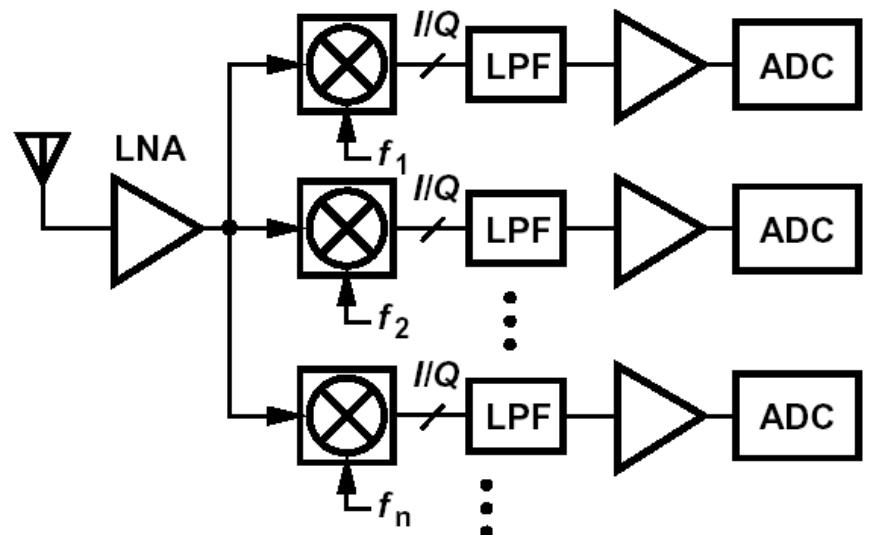
Commur

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UCLA

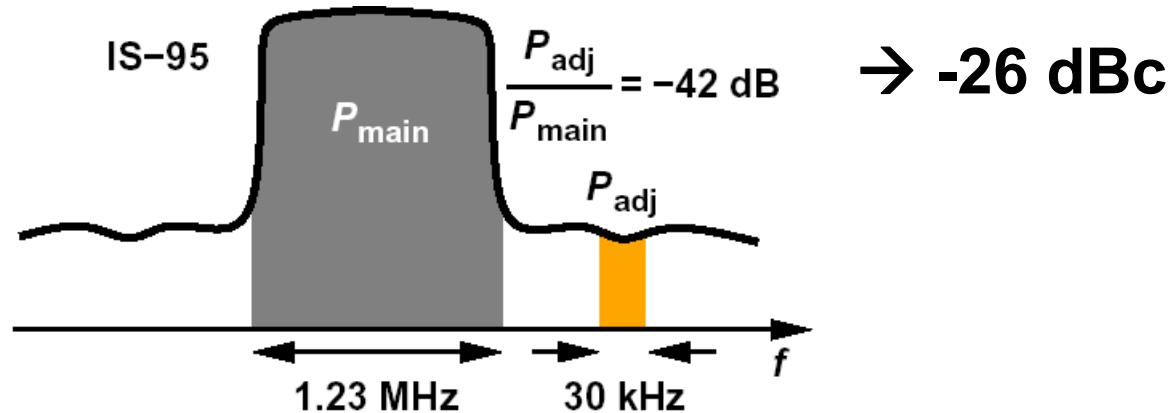
# RX Path

- Broadband gain and input matching
  - Difficult to switch different circuits in and out at the input.
- Low noise – especially flicker noise for 20-50 MHz
- High IP3 and IP2
- Multiple concurrent downconversions to speed up spectrum sensing:



# TX Path

- Broadband upconversion, PA, and matching
- Low adjacent-channel power



- Concurrent Transmission and Sensing?
- Concurrent Transmission in multiple channels?



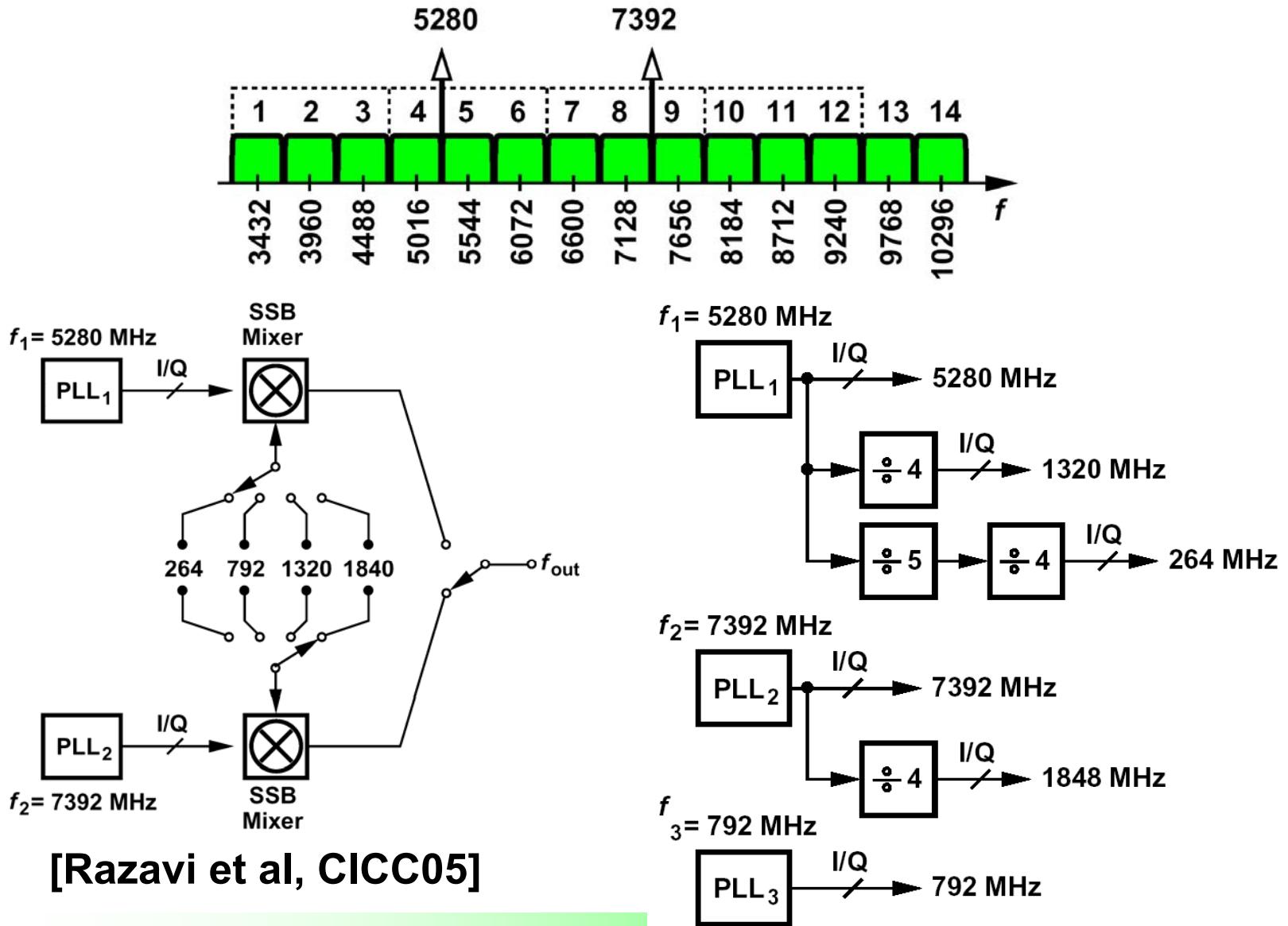
# Frequency Synthesis: 2.5 Decades

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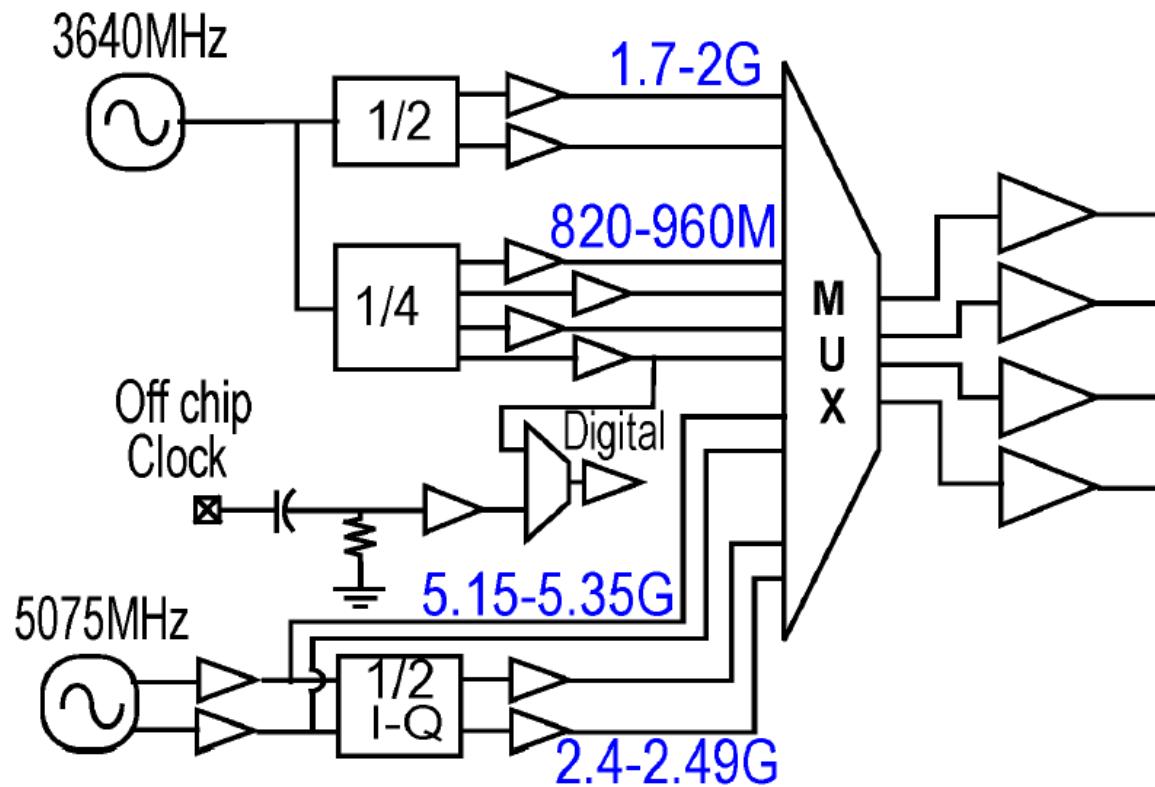
- LC VCO Tuning Range  $< +/- 10\%$
- If a frequency is divided by an odd number, it must then be divided by 4 to generate quadrature phases.
- Single-sideband mixing probably out of the question
- How many VCOs does it take to cover one decade?



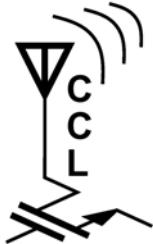
# UWB Example



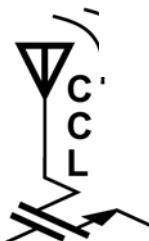
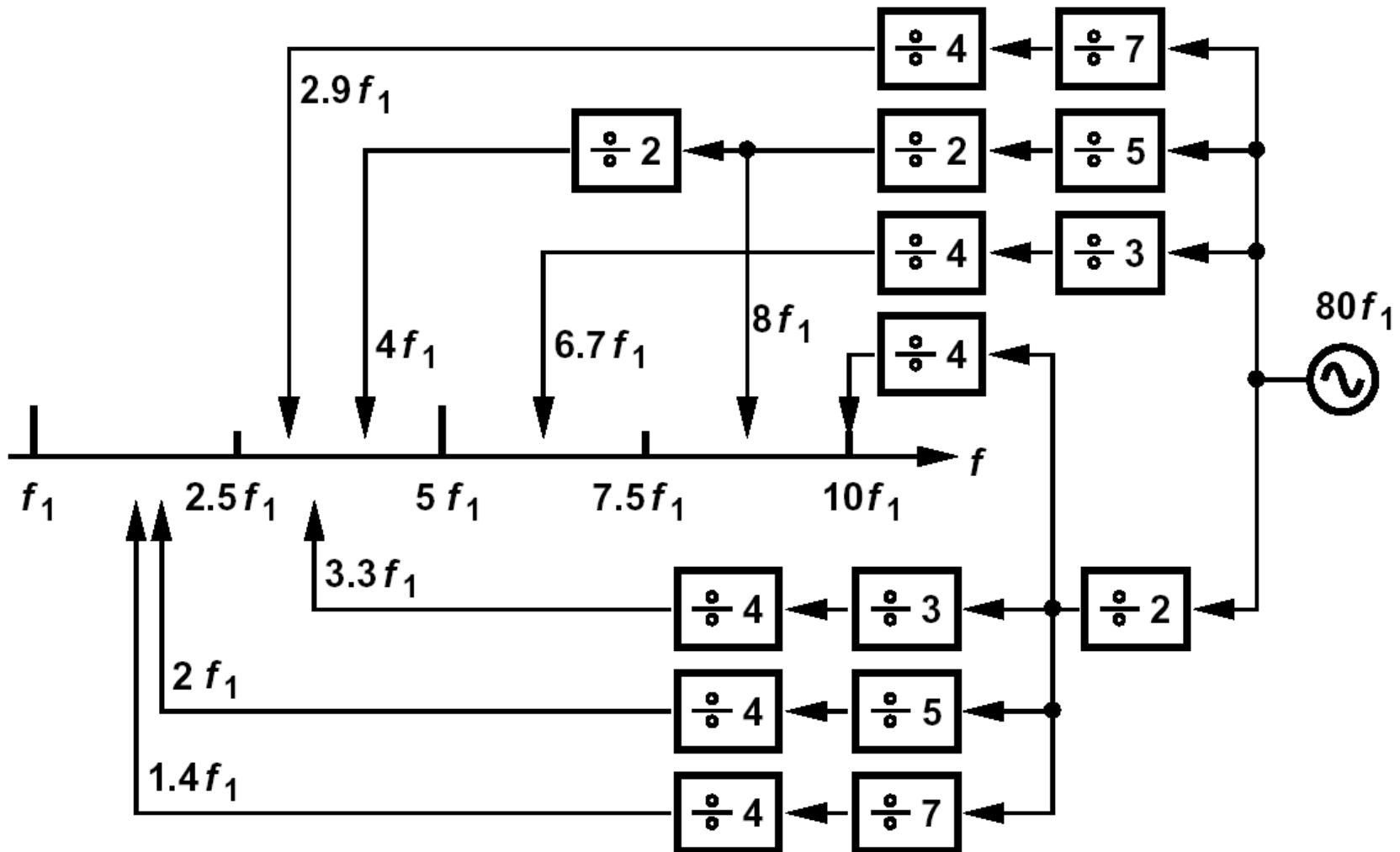
# SDR Example



[Bagheri et al, ISSCC06]

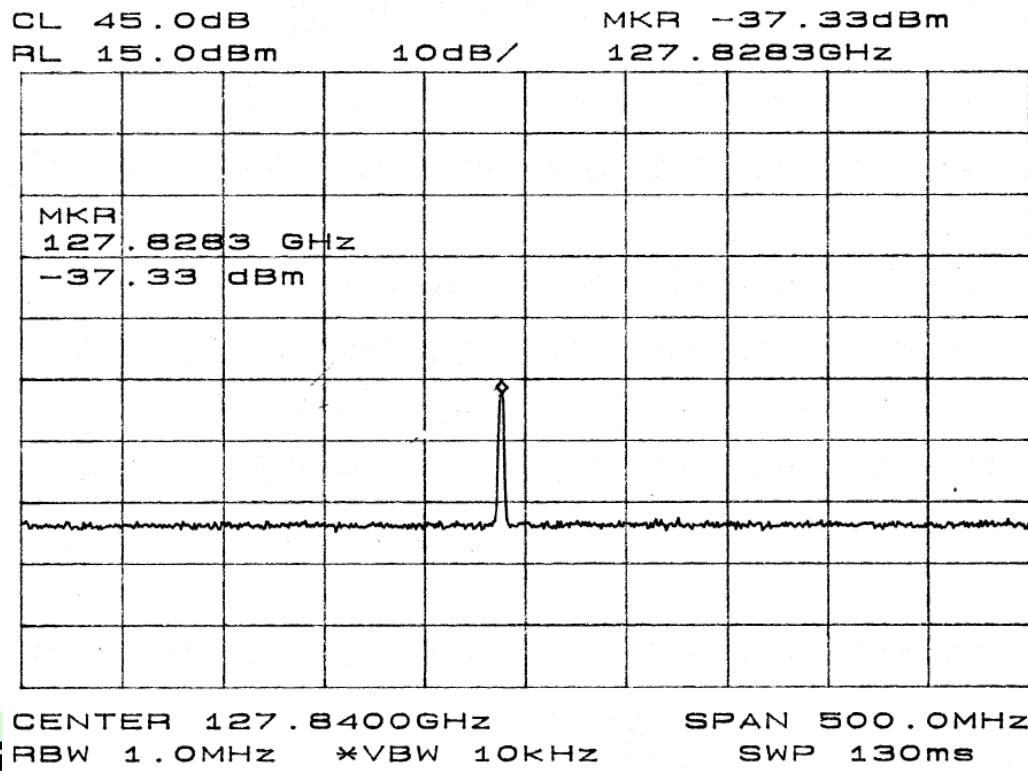
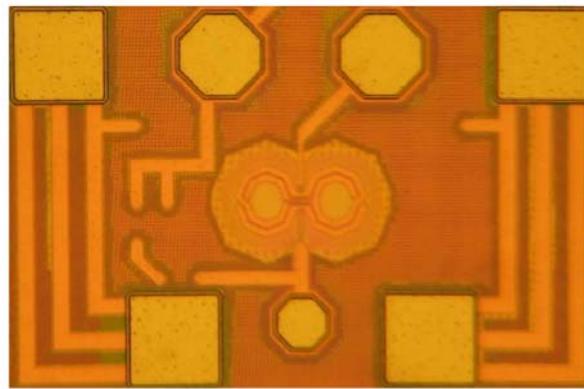
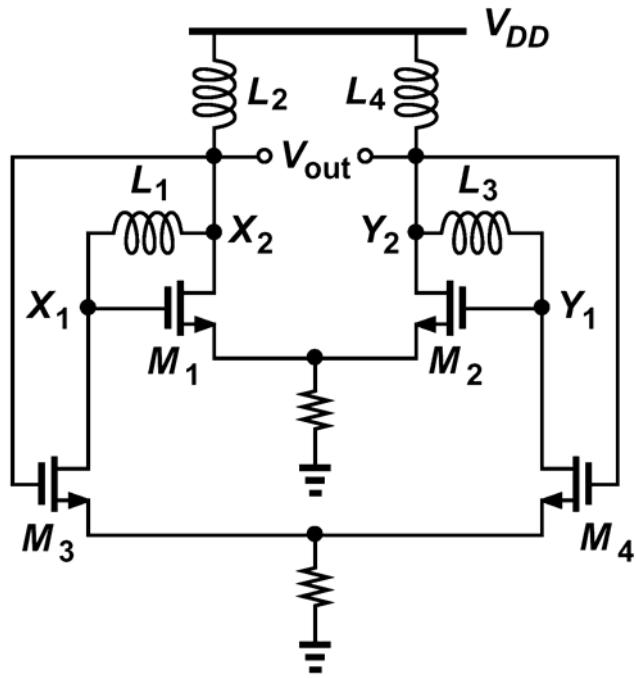


# How to Cover One Decade?



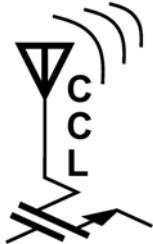
# Millimeter Waves to the Rescue

## 128-GHz Osc. in 90-nm CMOS



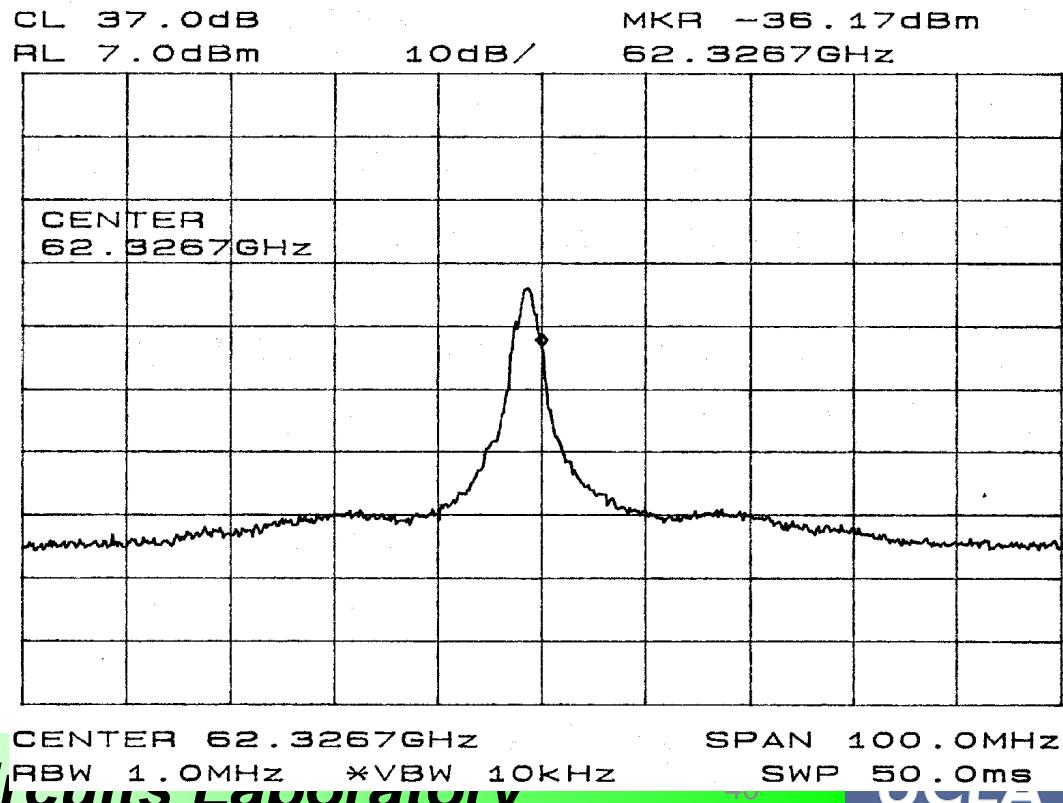
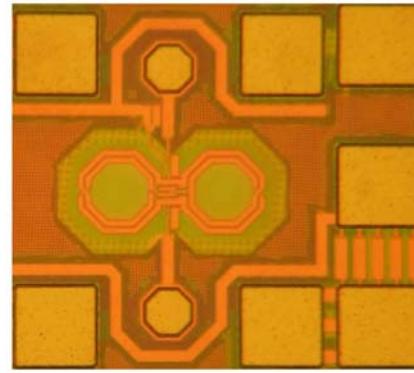
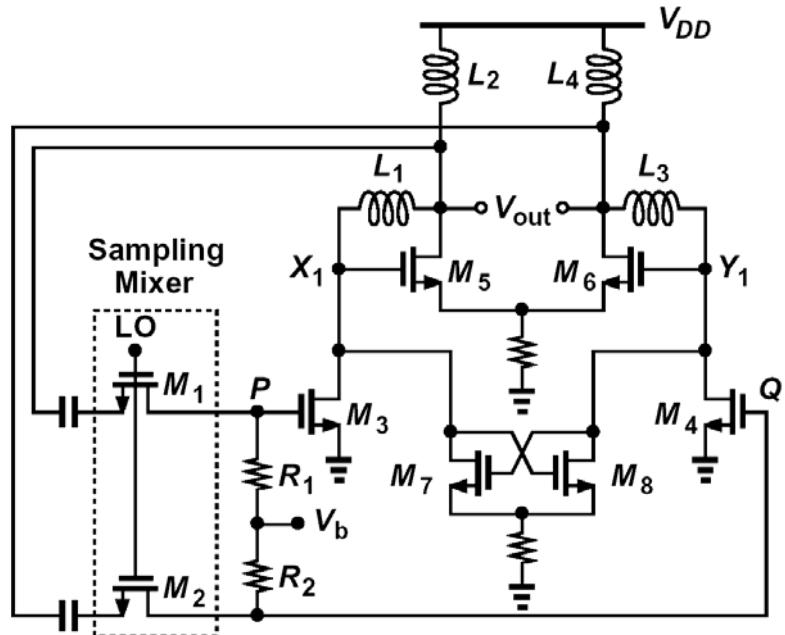
[Razavi, JSSC, Sept. 08]

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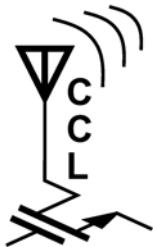


# Millimeter Waves to the Rescue

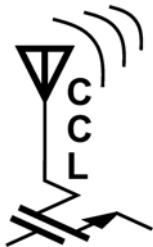
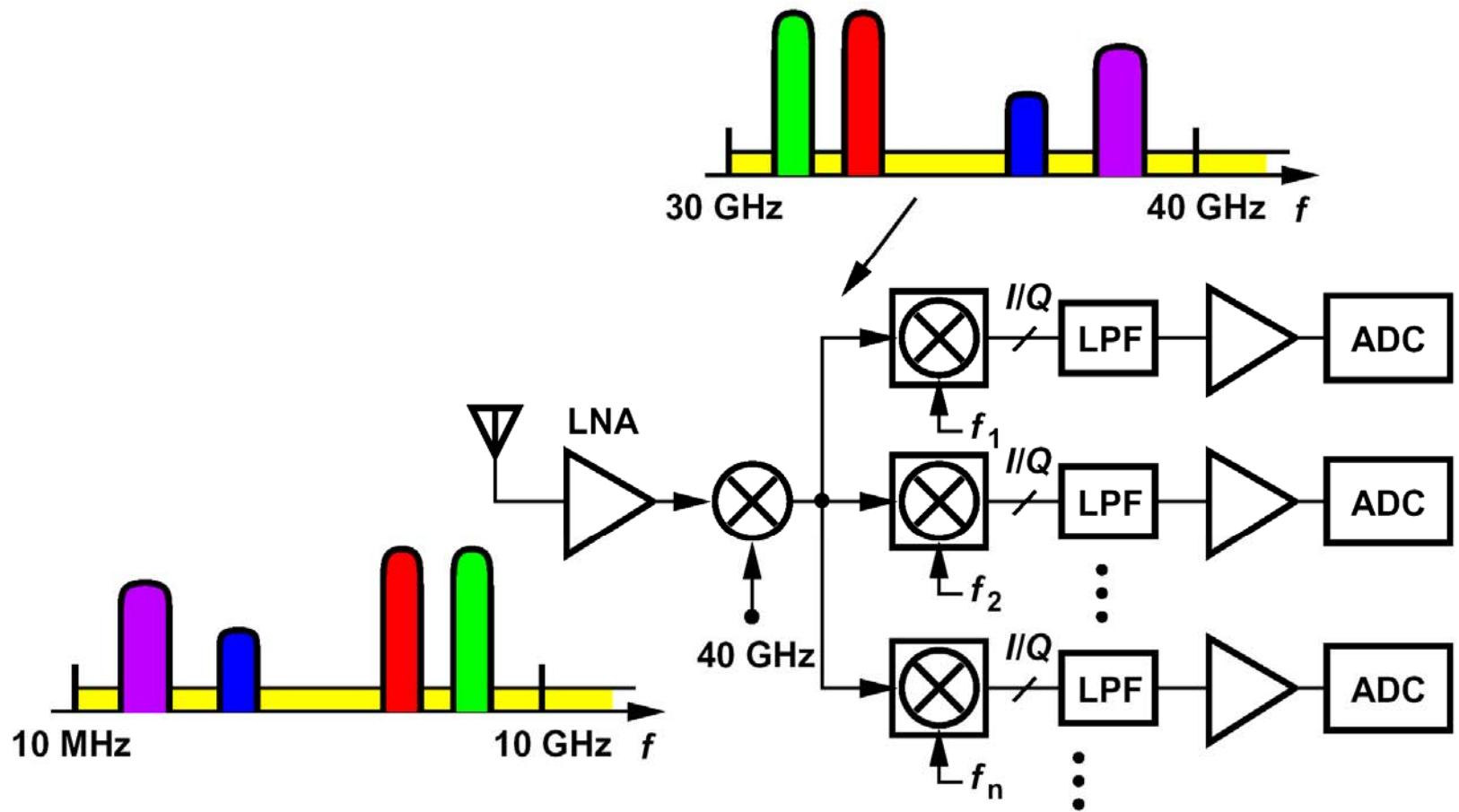
## 125-GHz Divider in 90-nm CMOS



[Razavi, JSSC, Sept. 08]



# Alternative Solution



# Conclusion

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- Millimeter-wave and cognitive radios pose new challenges in RF and analog design – thereby keeping us employed.
- Cross-fertilization of concepts from UWB and mm-waves can greatly benefit CRE design.
- Many issues need to be studied and quantified:
  - Baseband ADC Requirements
  - NF Calibration
  - Coverage of 2-3 Frequency Decades
  - Broadband Gain, Matching, PAs, etc.

