

# ***Radio Research Directions***

**Behzad Razavi**

**Communication Circuits Laboratory  
Electrical Engineering Department  
University of California, Los Angeles**



# Outline

---

- Introduction
- Millimeter-Wave Transceivers
  - Applications
  - Challenges
  - Examples
- Cognitive Radios
  - Challenges
- Conclusion



# Nature of Our Radio Research

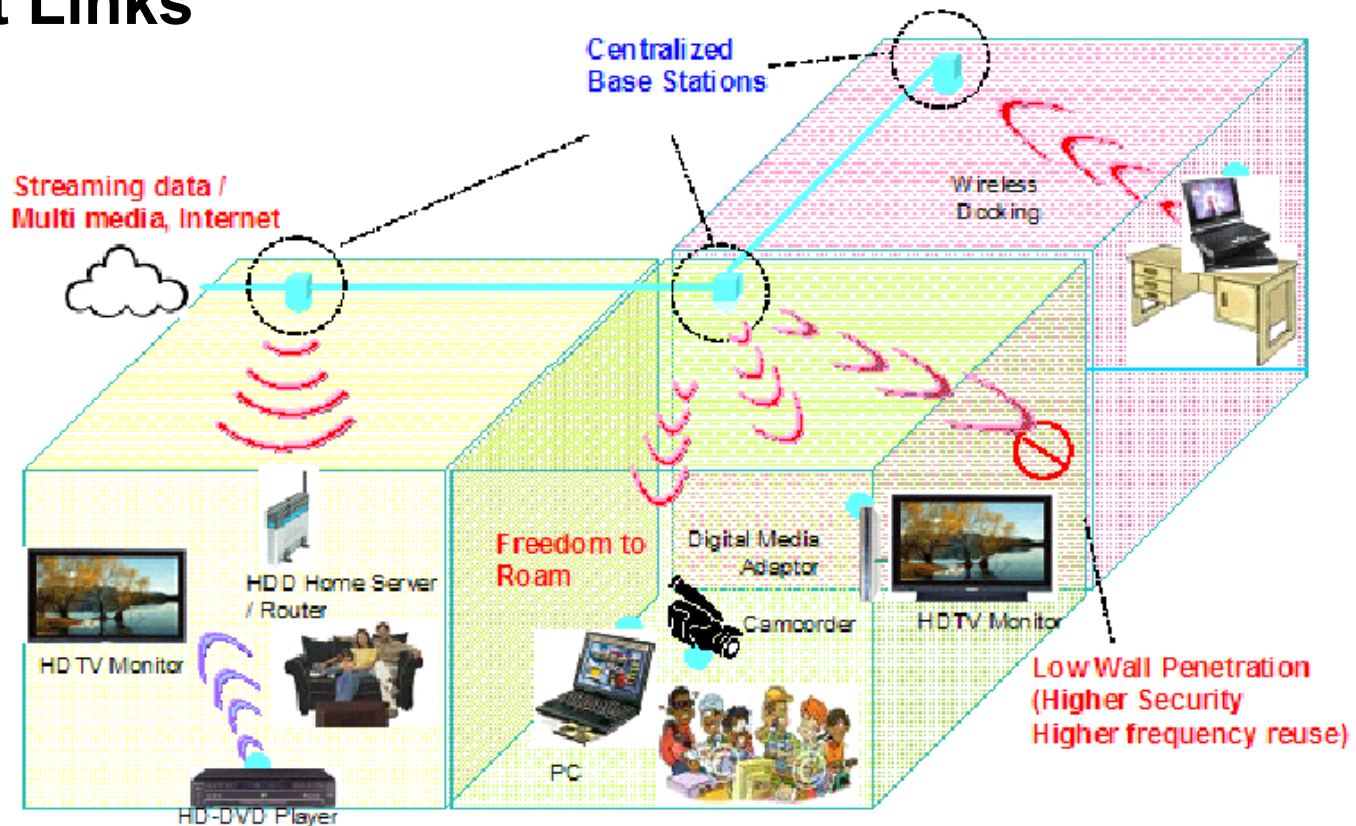
---

- Target highly-integrated transceivers with minimal number of external components
- Address tough problems: higher frequency, wider bandwidth, lower power consumption, ...
- Develop new architectures, circuits, devices that solve these problems
- Realize ideas in standard CMOS technology and verify by experimentation
- Examples of past work:
  - 900-MHz/1.8-GHz Transceivers for Cellular Telephony
  - 2.4-GHz and 5.2-GHz WLAN Transceivers
  - 3-6 GHz UWB Transceivers
  - 5-GHz RX for MIMO
  - 60-GHz Transceivers



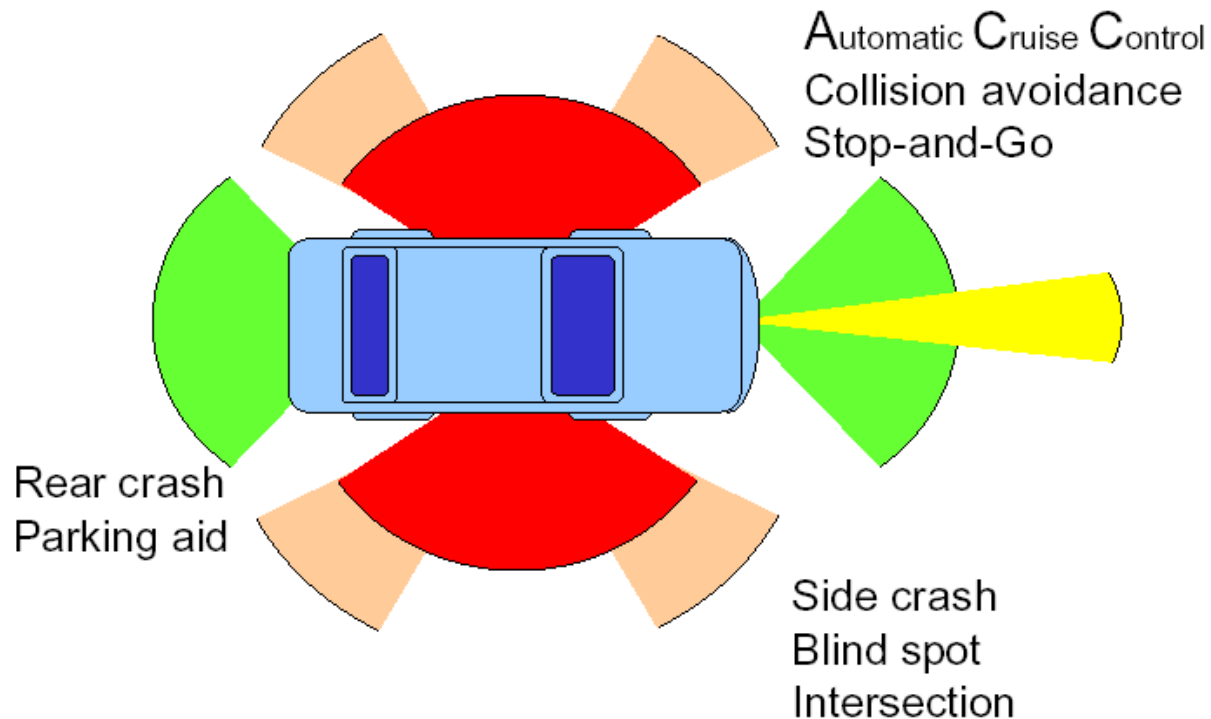
# Why the interest in mmWaves?

- **Unlicensed band: 57 GHz – 64 GHz offers possibility of high-data rate communications:**
  - **High-Definition Video Streaming**
  - **Fast Links**



# Why the interest in mmWaves?

- Automotive Radar (60-77 GHz)



# Why the interest in mmWaves?

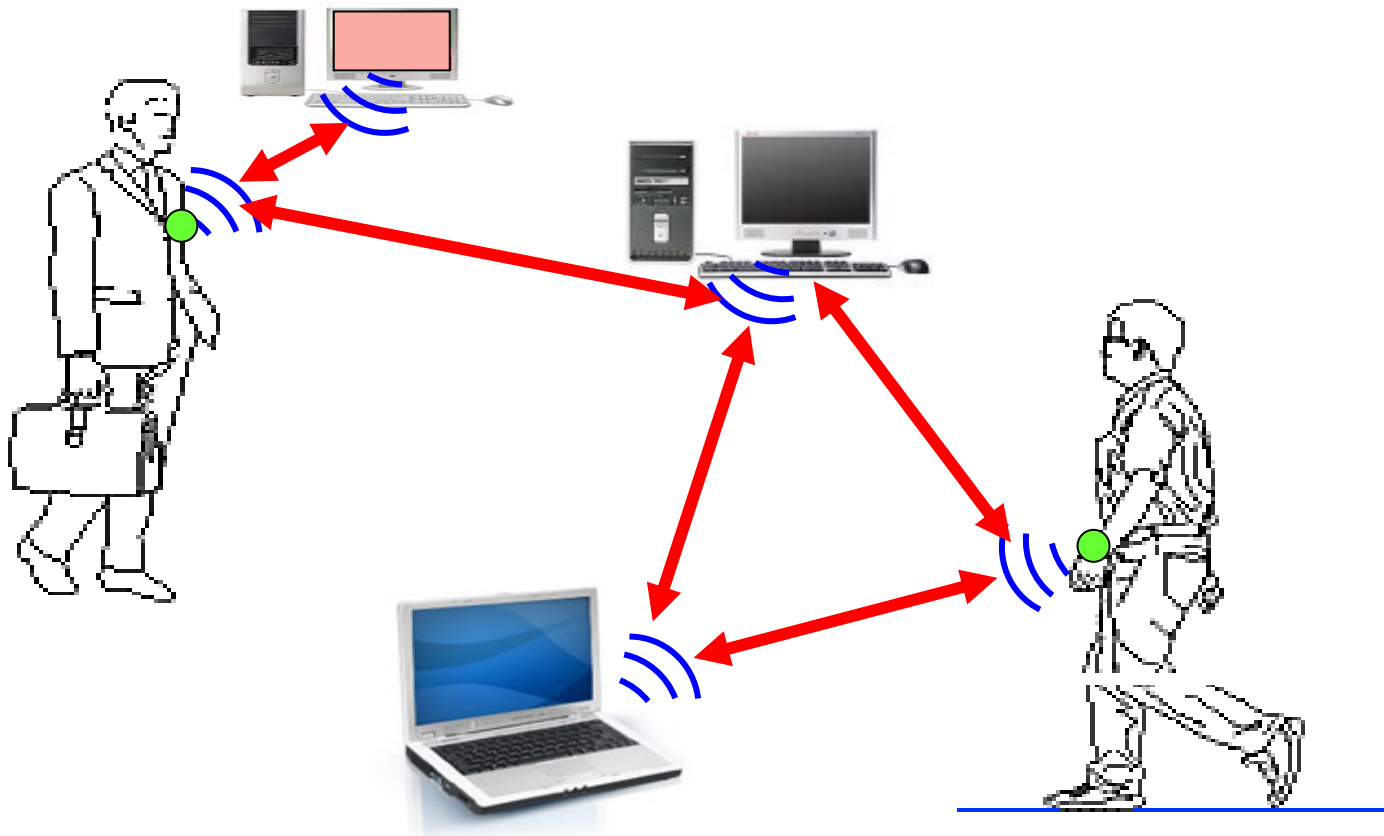
---

- mmWave Imaging ( $> 100\text{GHz}$ )

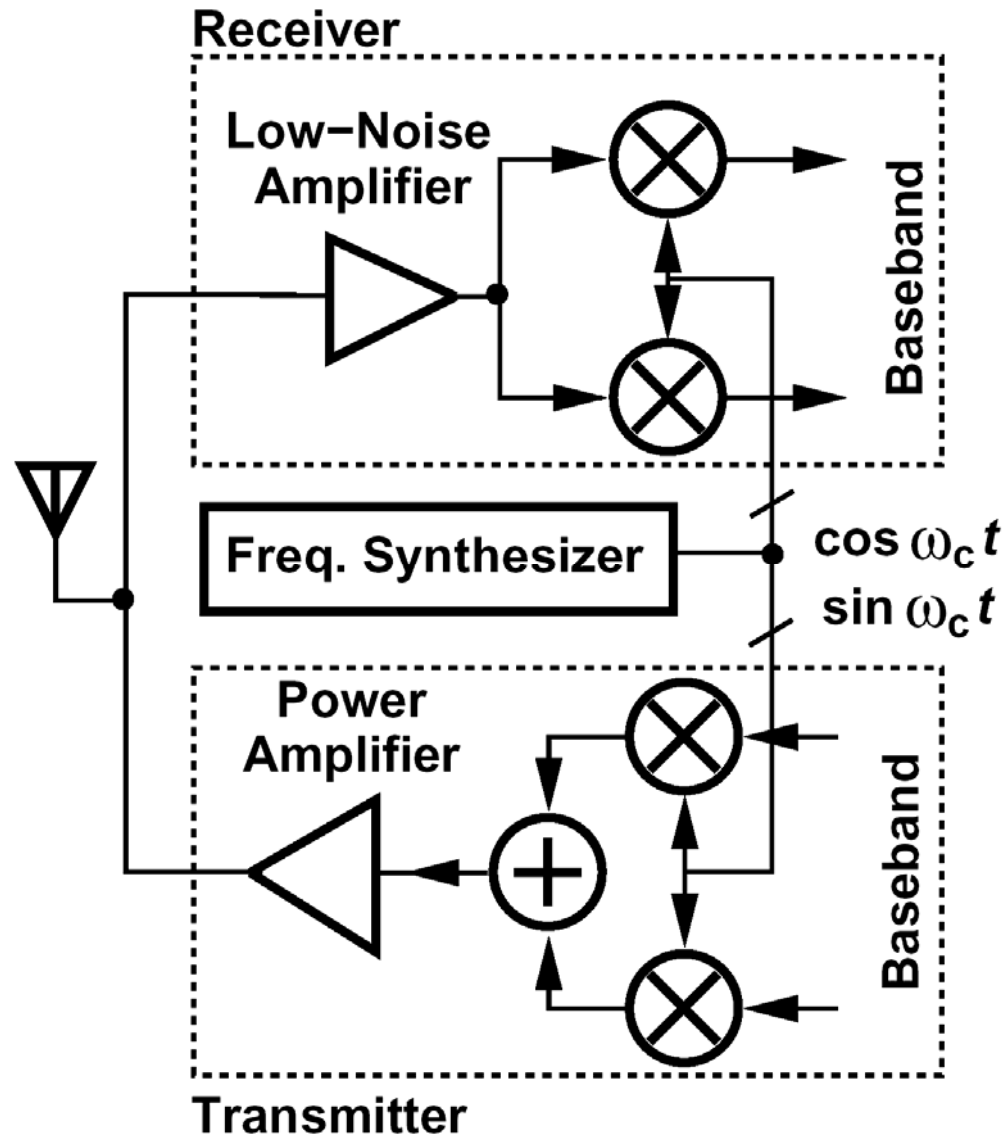


# Networks with High Redundancy

- Line-of-sight propagation a serious issue

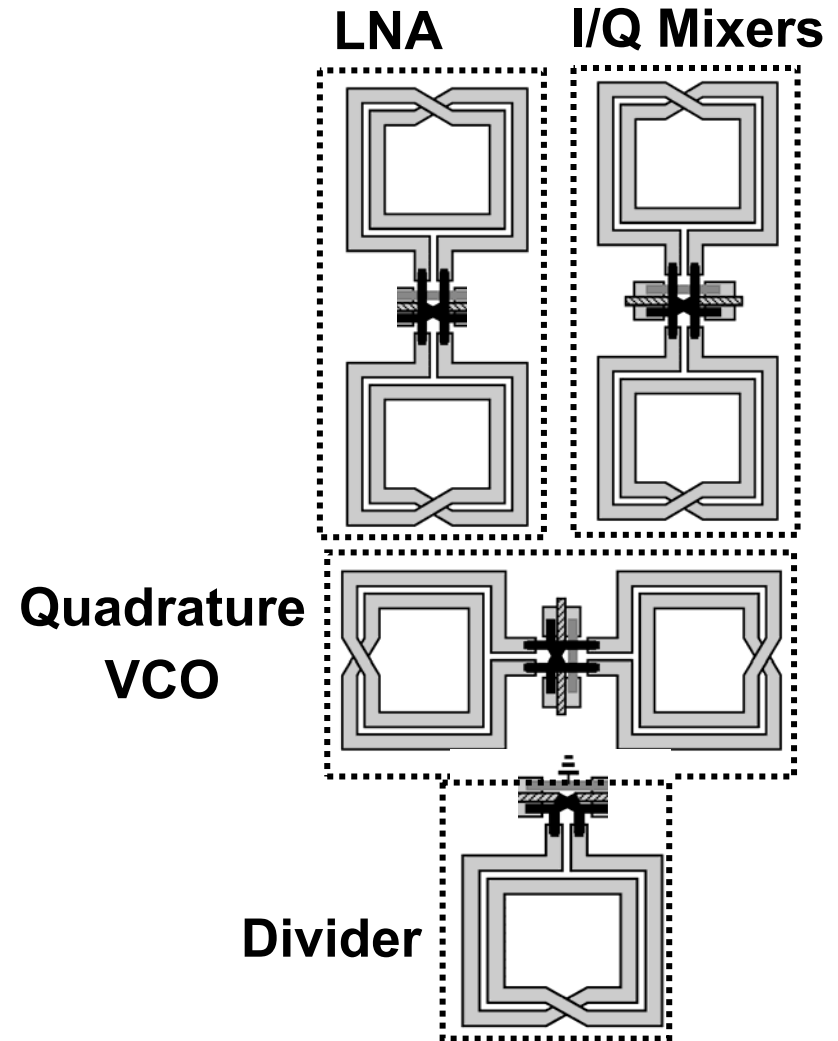
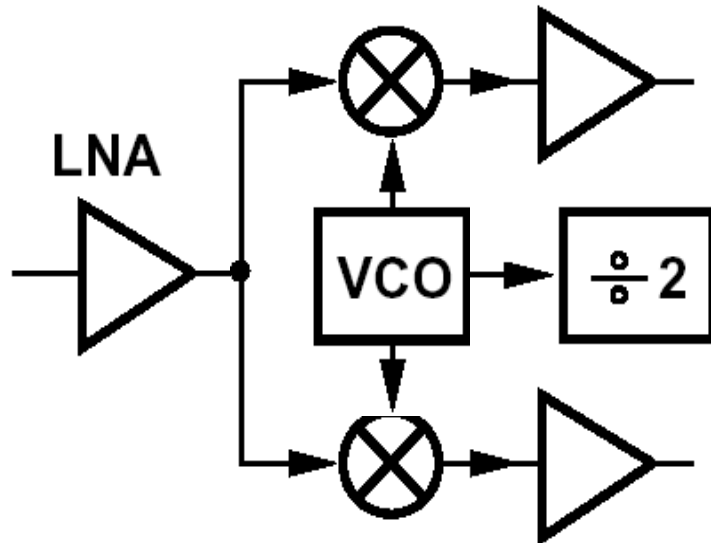


# A Few Words for the RF-Challenged



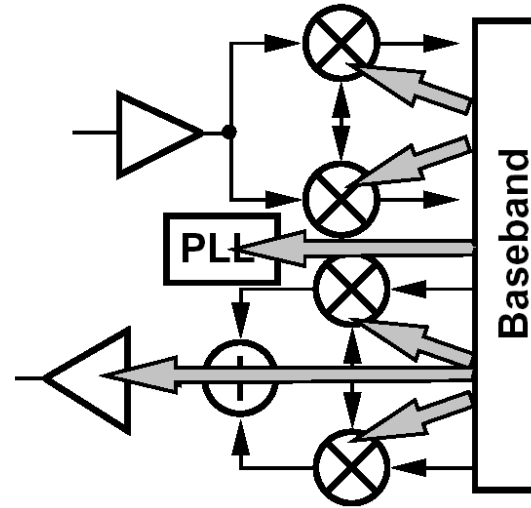
# Architecture-Level Challenges

- LO (I/Q) Generation
- LO Division
- LO Distribution

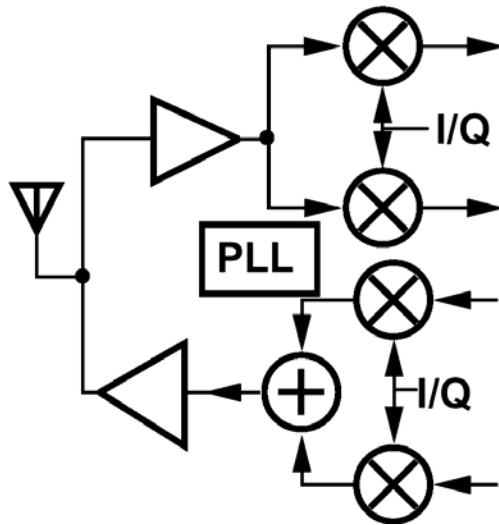


# Innovation at All Levels

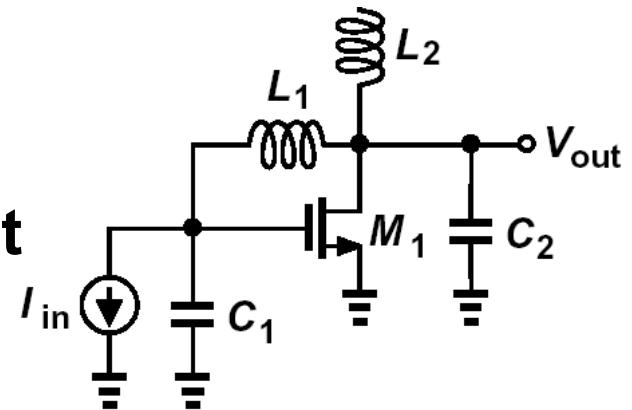
**System**



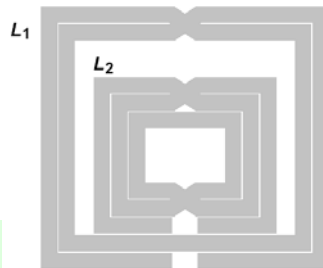
**Architecture**



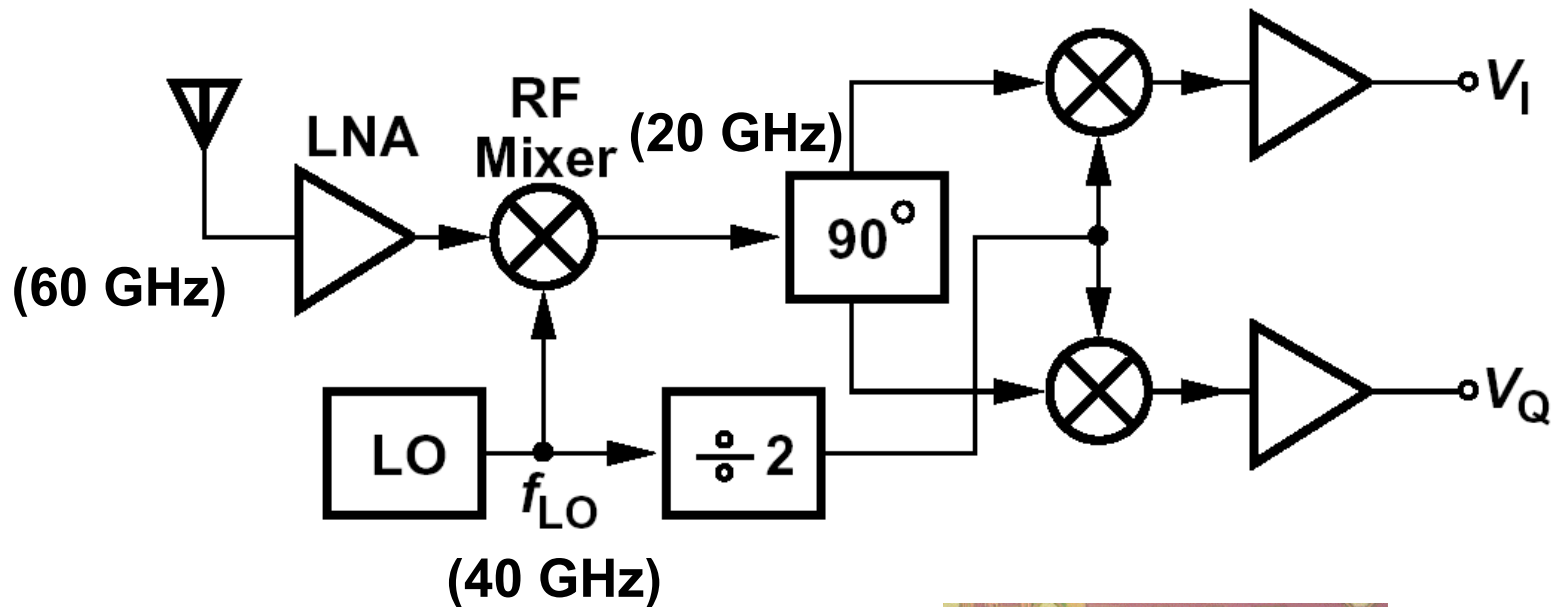
**Circuit**



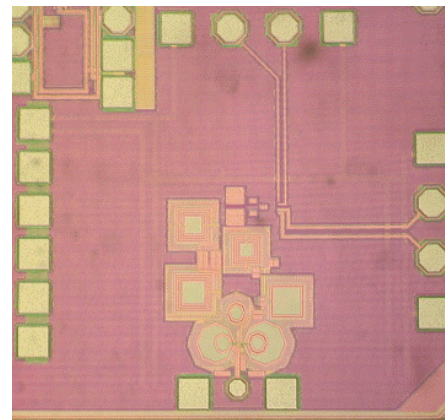
**Device**



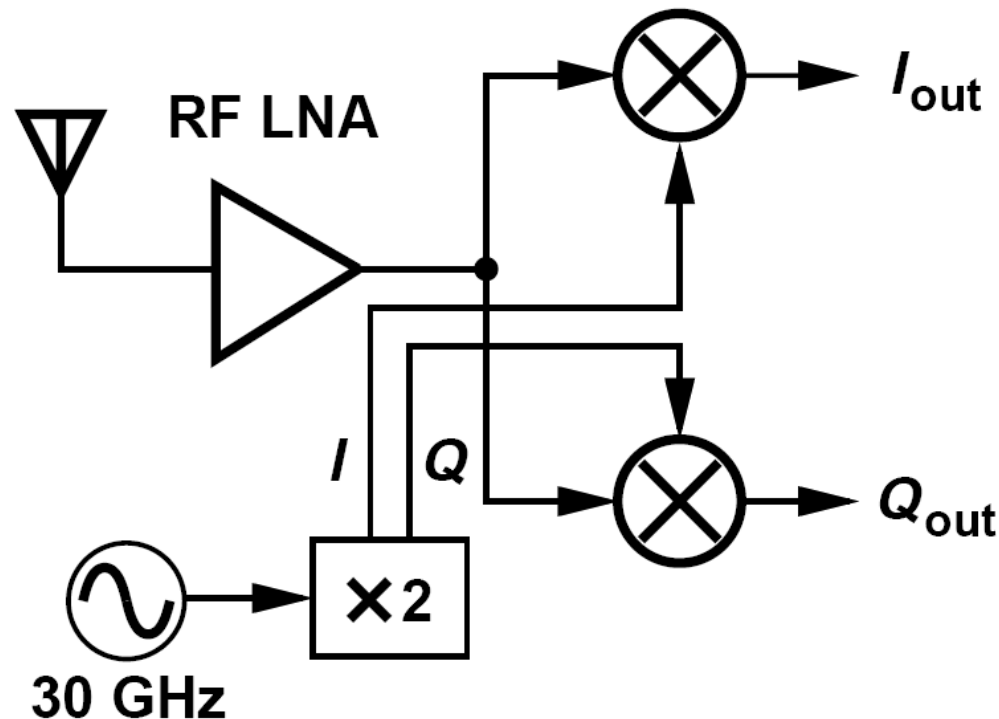
# Our Second-Generation 60-GHz RX



[B. Razavi, ISSCC 07]



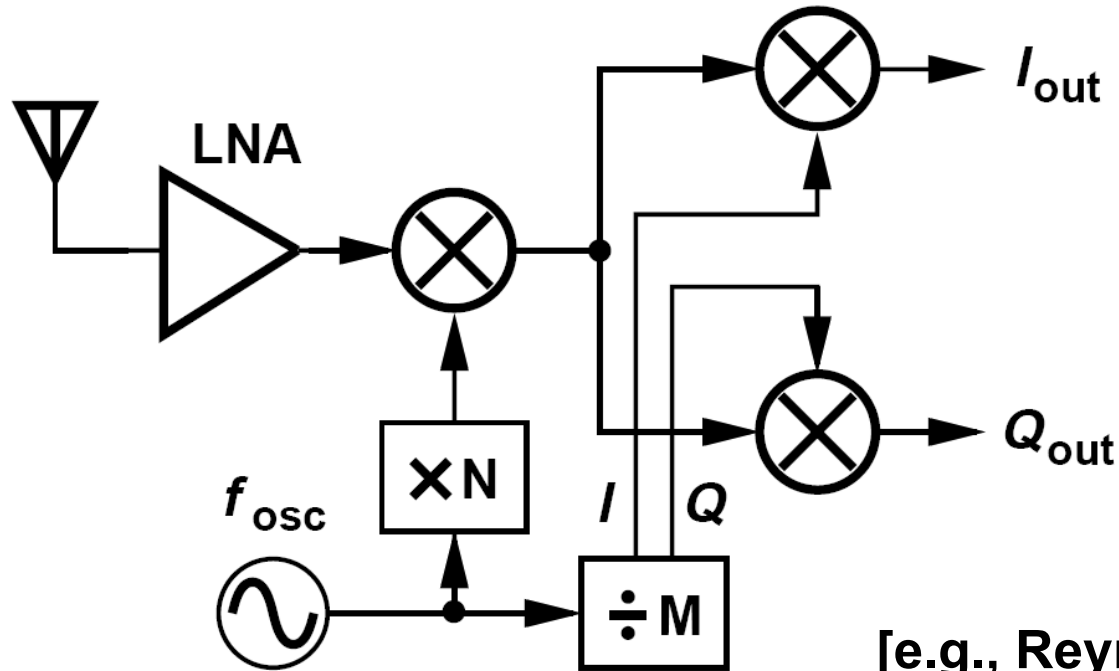
# Direct-Conversion RX with 30-GHz LO?



- Quadrature generation is difficult.
  - Distribution is difficult.
- Need “synthesizer-friendly” transceivers.



# Heterodyne Receiver

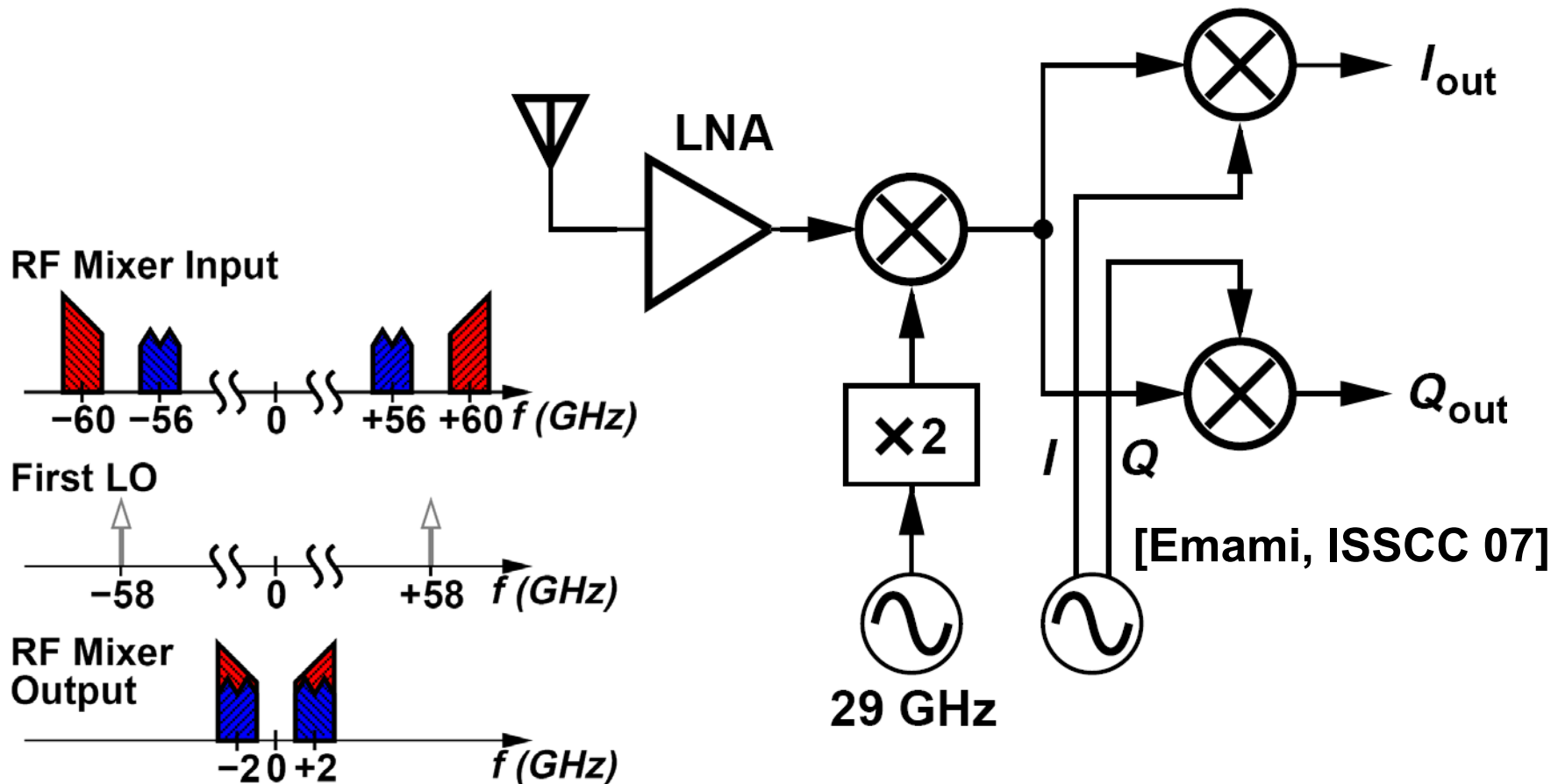


[e.g., Reynolds, JSSC, Dec 06]

- Multiplier has high loss and needs its own inductor.



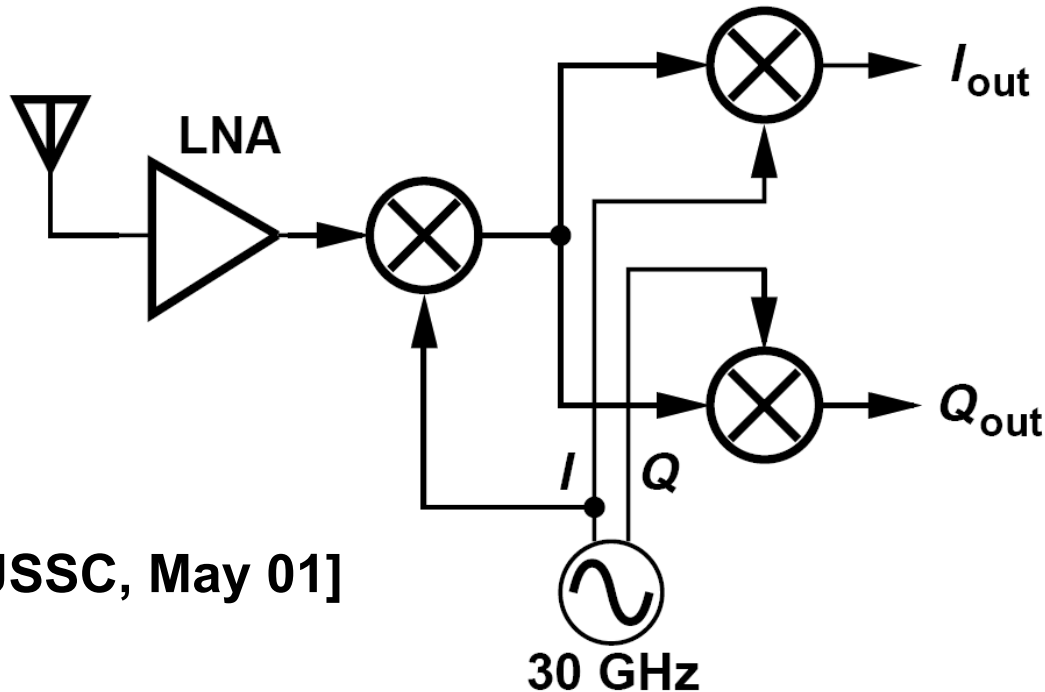
# Problem of Low-IF Heterodyne



- Image of the first mixer is in the band.
- Receiver NF is increased by  $\sim 3$  dB.



# Example of Synthesizer-Friendly Receiver



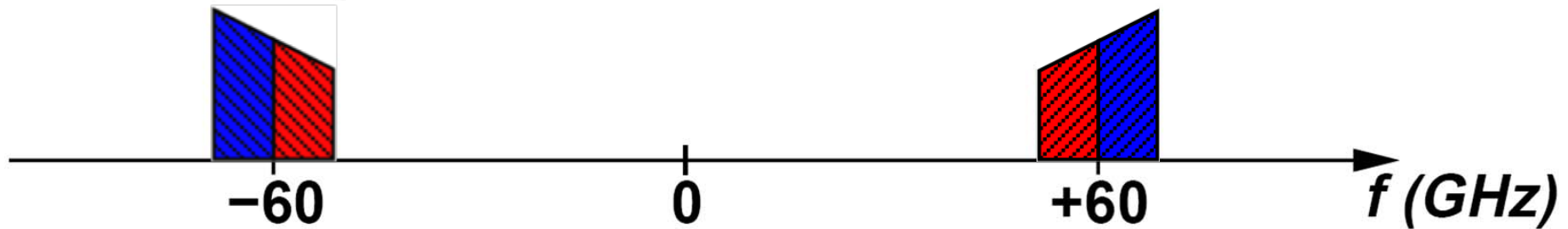
[Razavi, JSSC, May 01]

- No extra divider/multiplier needed.
- Image is at DC.
- But,
  - Third harmonic of LO causes corruption.
  - LO-IF feedthrough may desensitize the IF mixers.
  - $1/f$  noise is upconverted in RF mixer.

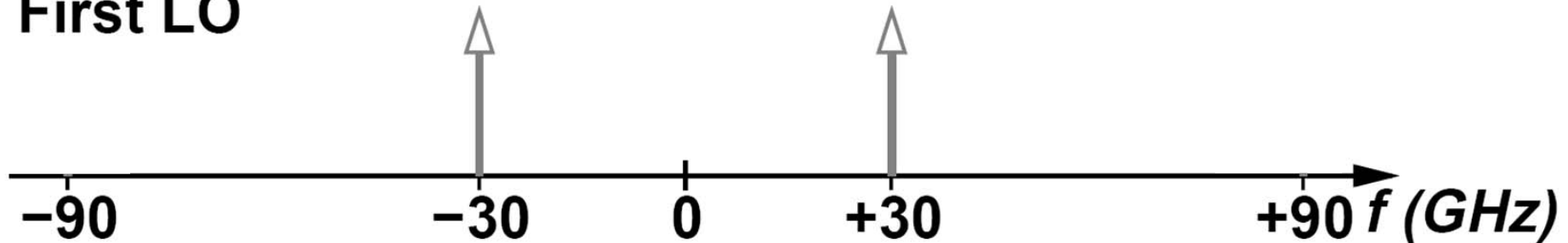


# Problem of LO Third Harmonic

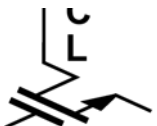
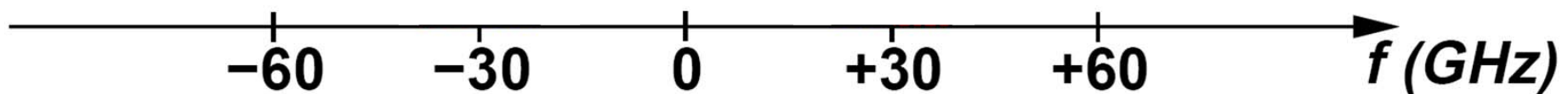
RF Mixer Input



First LO



RF Mixer Output



# Analysis

$$x_{RF}(t) = \Re \left[ x_{BB}(t) e^{+j2\pi f_{RF}t} \right]$$

$$LO = \cos \omega_{LO}t + \alpha \cos 3\omega_{LO}t \quad \alpha \approx 1/3$$

$$x_{IF}(t) = \Re \left[ \frac{x_{BB}(t) + \alpha x_{BB}^*(t)}{2} e^{+j2\pi f_{IF}t} \right]$$

$$x_{out}(t) = k \left[ x_{BB}(t) + \alpha x_{BB}^*(t) \right]$$

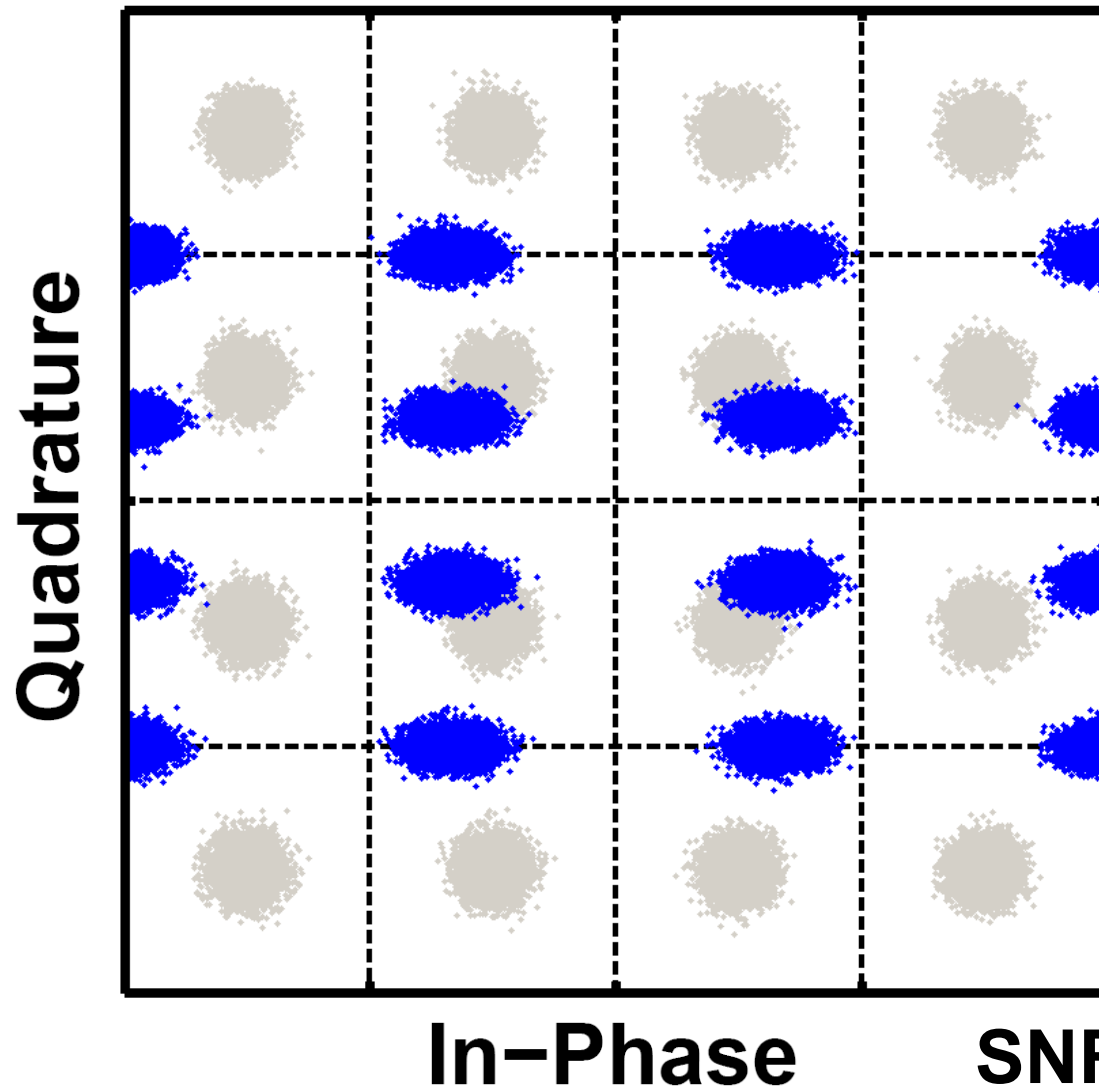
$x_{BB}(t)$  : **wanted signal**       $\mathcal{F} \{ x^*(t) \} = X^*(-f)$

$x_{BB}^*(t)$ : **mirrored replica of the signal**



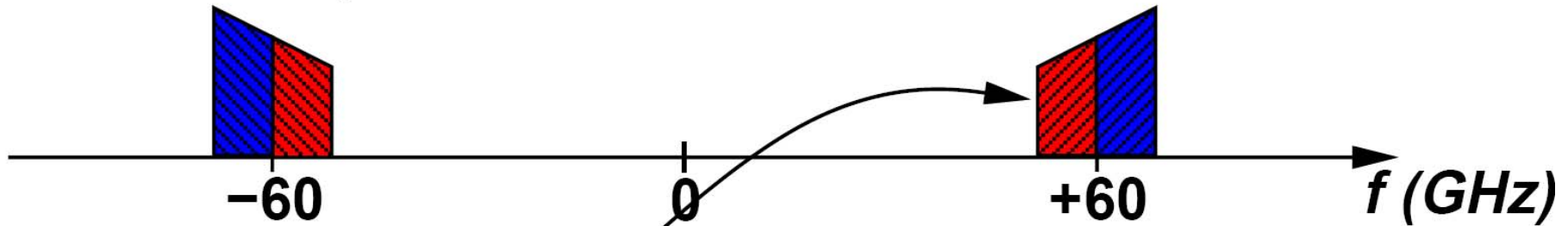
# 16-QAM Constellation

---

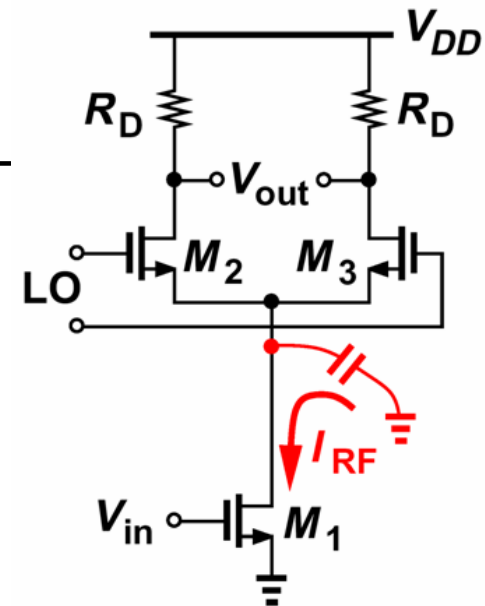
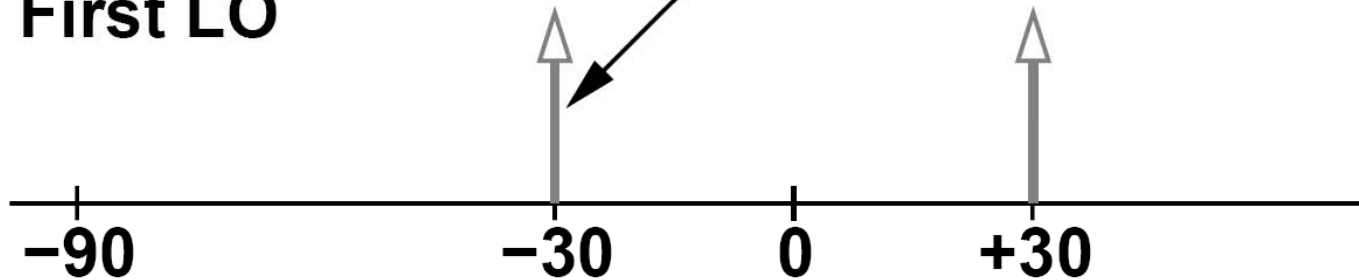


# Linearize LO Port?

RF Mixer Input

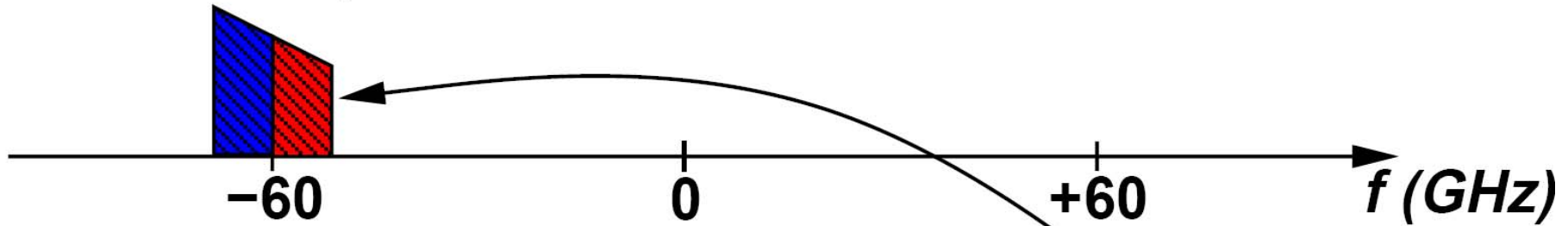


First LO

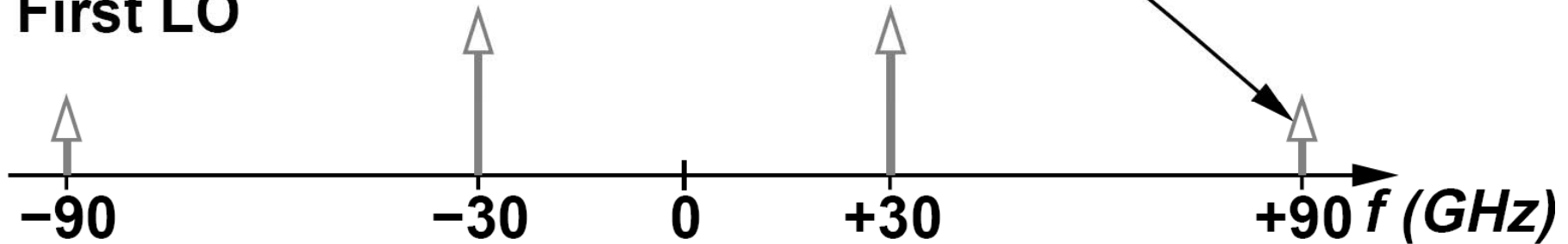


# Alternative Solution

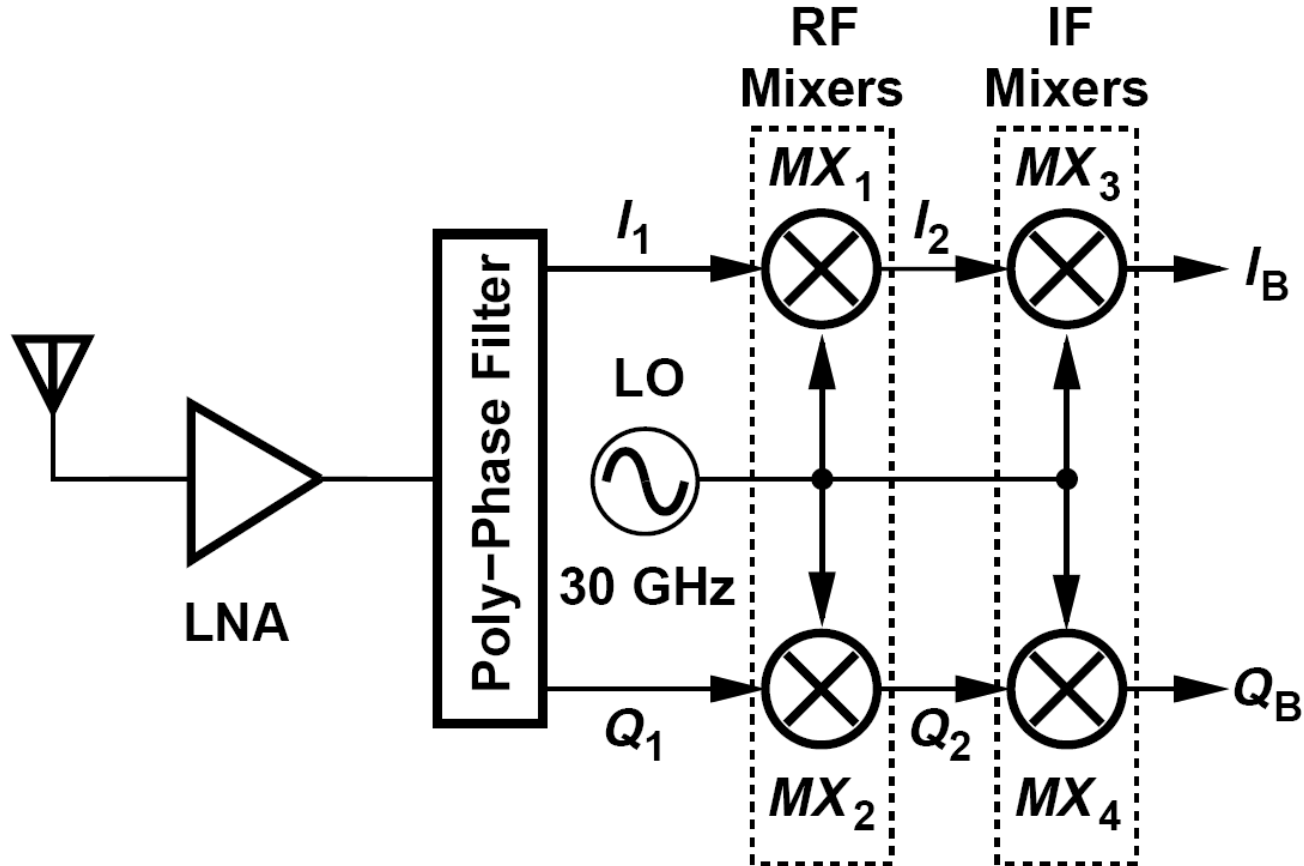
RF Mixer Input



First LO



# Proposed Receiver Architecture



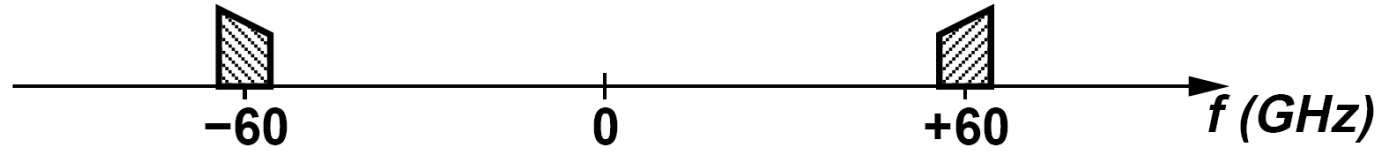
- Lowest possible LO frequency (without multiplication).
- No quadrature LO phases required.



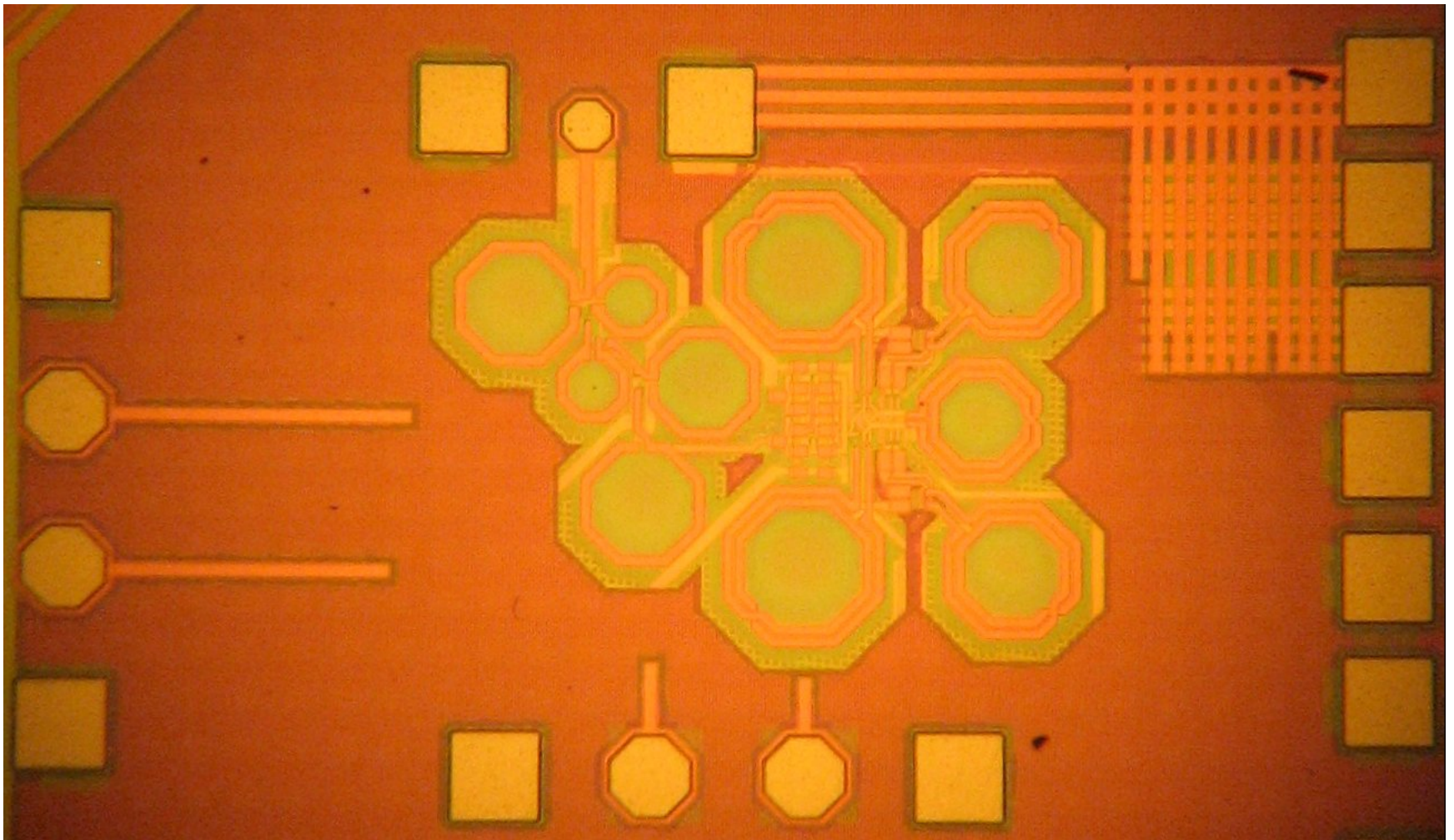
# Receiver Spectra



RF input



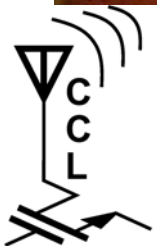
# Die Photograph



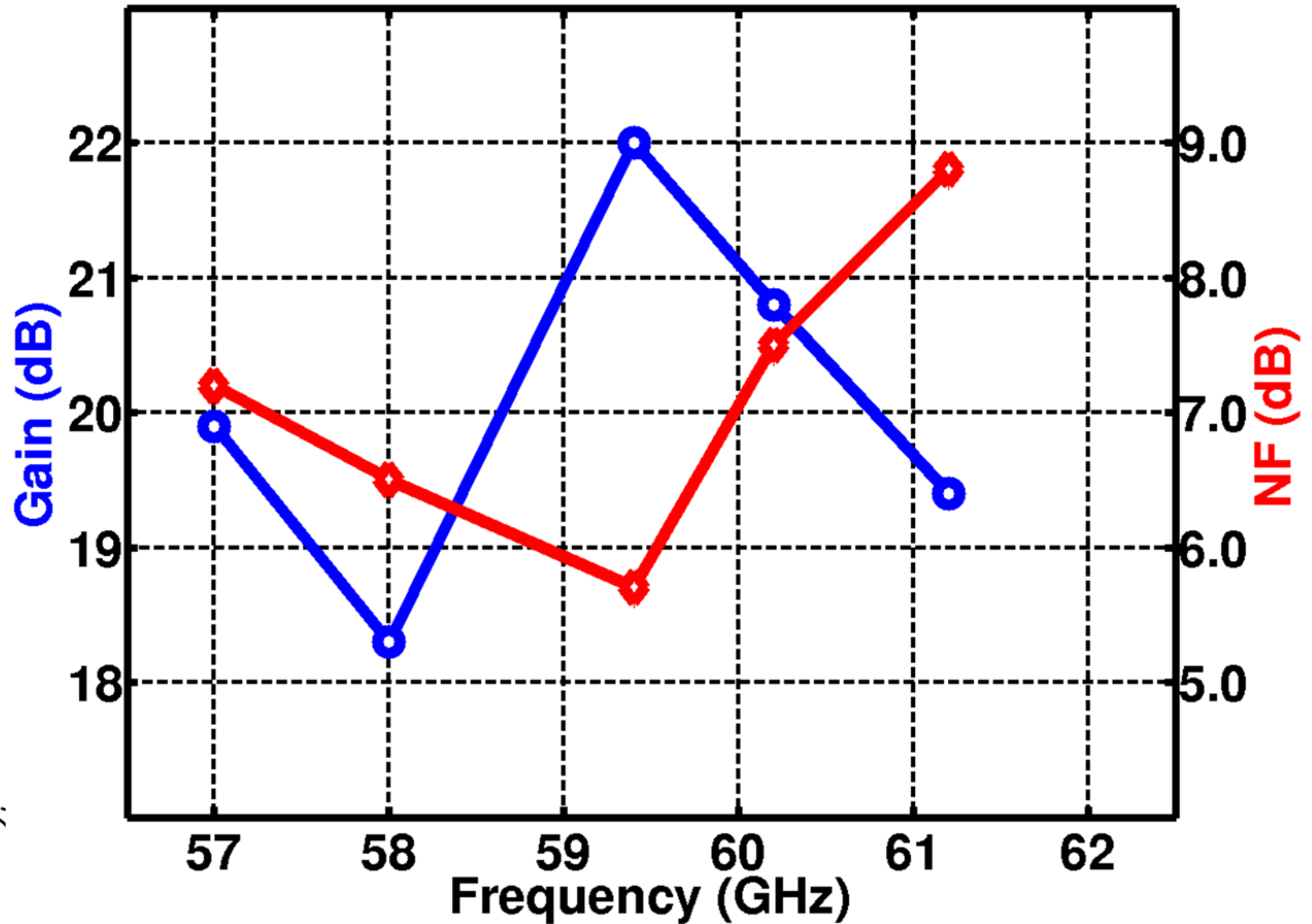
**Fabricated in TSMC's 90-nm CMOS technology.**

**Active area: 500  $\mu\text{m}$  x 370  $\mu\text{m}$**

***Communication Circuits Laboratory***



## Measured NF and Gain



# Comparison

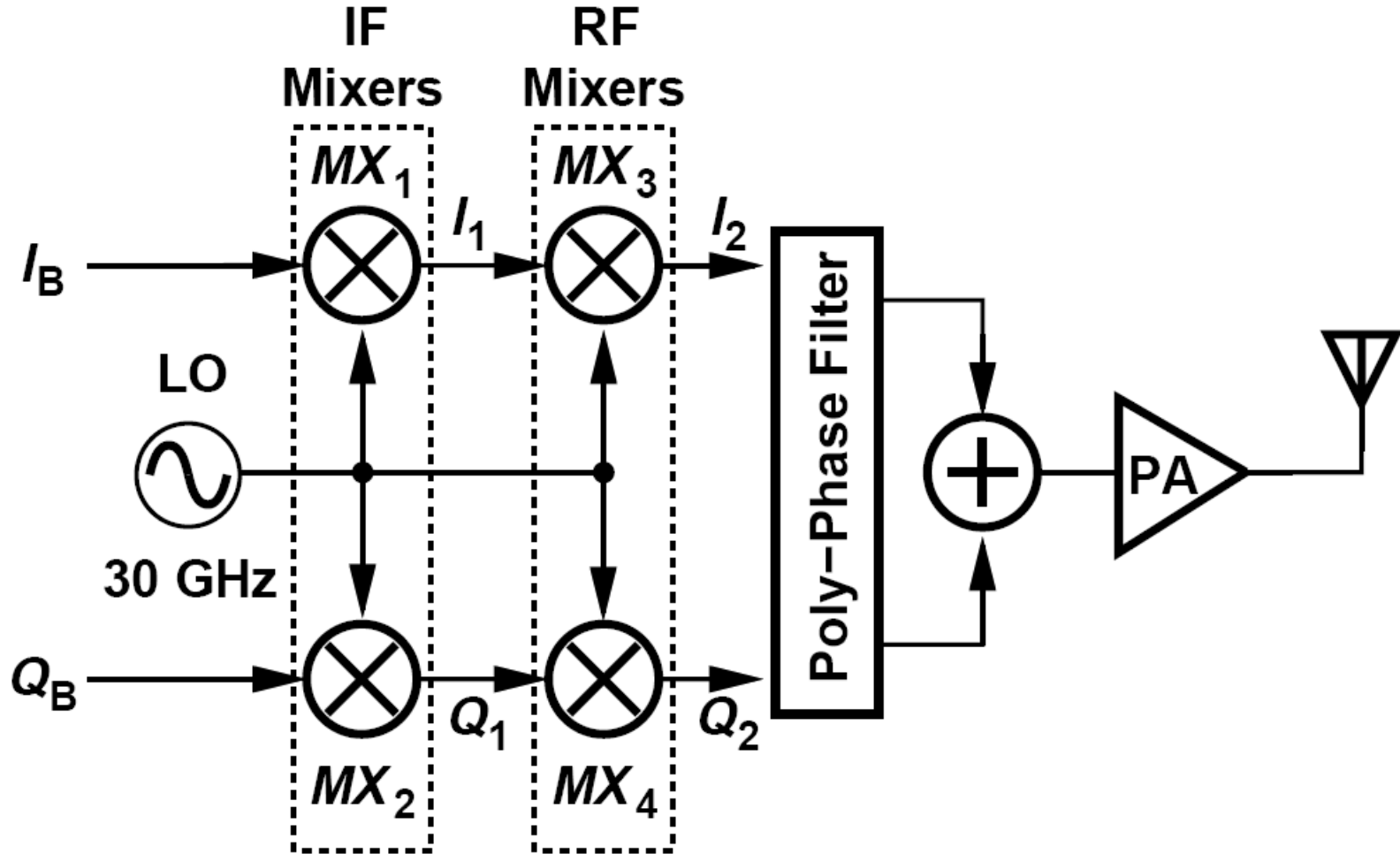
	Receiver in [3]	Receiver in [1]	This work
Noise Figure (dB)	10.4-11	6.9-8.3	5.7-8.8
Gain (dB)	9.5-12	26-31.5	18.3-22
P <sub>1dB</sub> (dBm)	-15.8	-25.5	-27.5
LO Leakage to Input (dBm)	N/A	-47	-65
I/Q Mismatch	N/A	6.5°/ 1.5dB	2.1°/ 1.1dB
LO Phase Noise (dBc/Hz @ 1-MHz offset)	-86	-95	-87
Power Dissipation (mW)	77	80	36
LNA			9
Mixers			23
Oscillator			4
Supply Voltage (V)	1.2	1.8	1.2
CMOS Technology	0.13- $\mu$ m	90-nm	90-nm

[1] B. Razavi, ISSCC '07

[3] S. Emami et al, ISSCC '07



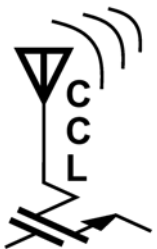
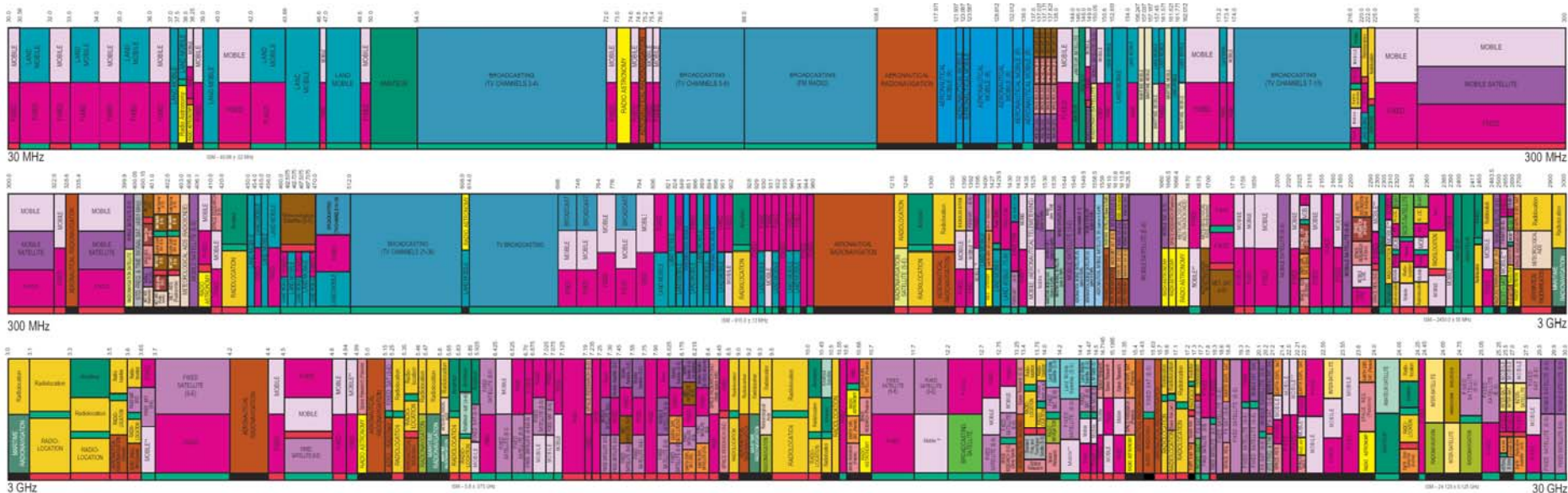
# Transmitter Architecture



- Does not require quadrature LO



## Detect and use unoccupied channels.



# RF/Analog PHY Design Issues

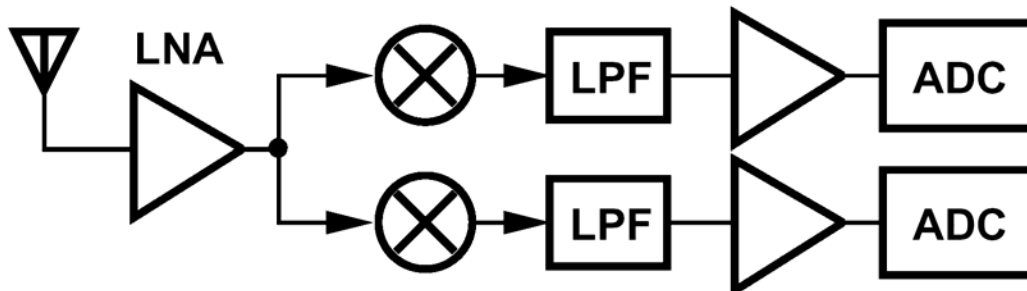
---

- Spectrum Sensing
- RX Path
- TX Path
- Frequency Synthesis



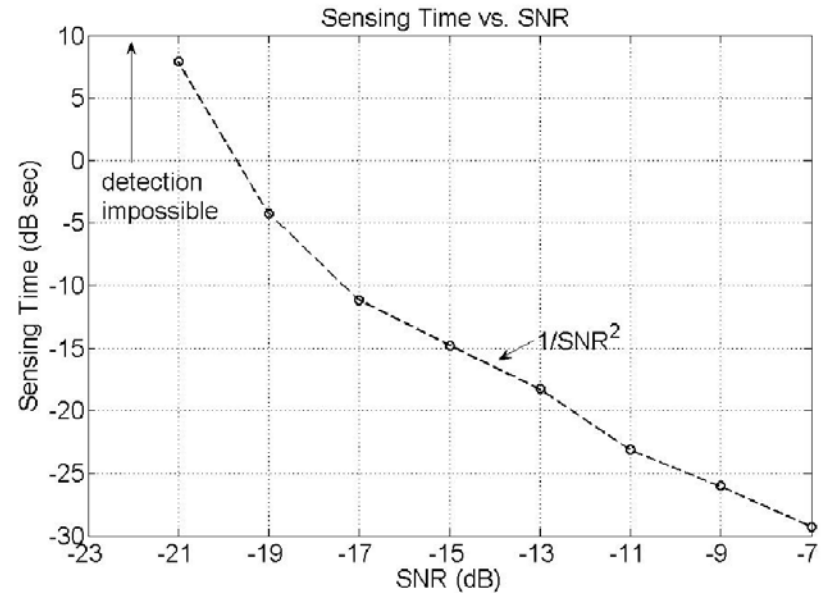
# Spectrum Sensing (I)

- Three Techniques:
  - Energy Detection
  - Pilot Detection
  - Signal Feature Detection
- Need to measure SNR  $\sim -20$  dB  $\rightarrow$ 
  - Accurate calibration of RX NF  
(i.e., need a tone with accurate amplitude)
  - Need enough gain to raise RX noise to well above 1 LSB of ADC



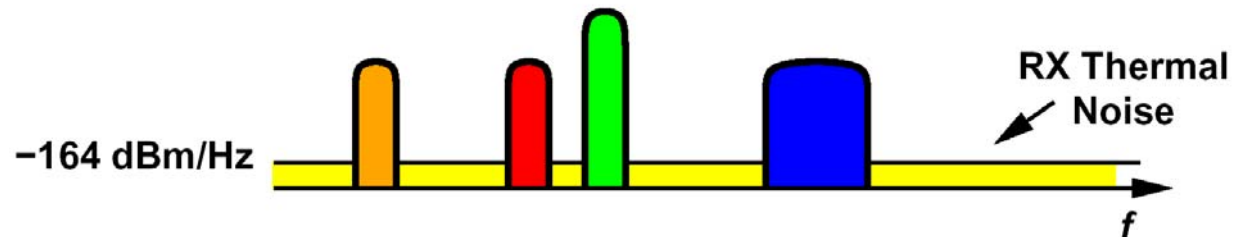
# Spectrum Sensing (II)

- **Channel-by-Channel Sensing**
  - Relaxed ADC design ( $\sim 3$  bits)
  - Takes forever.  
(e.g., 4-MHz QPSK channel:  
30 ms for  $\text{SNR} = -17$  dB)
  - May not know the center or bandwidth of channel.



[Cabric, PhD Diss., UCB]

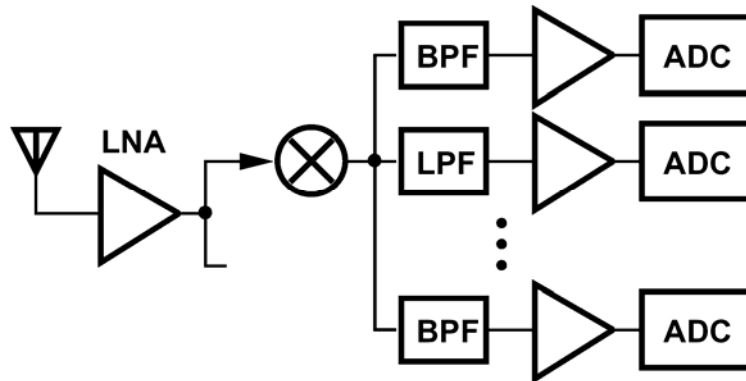
- **Block Downconversion Sensing**
  - Proportionally faster
  - But ADC BW and resolution much tougher



# Spectrum Sensing (III)

- **Two-Step Sensing:**

1. ADC takes a snapshot of a block of channels and determines “potentially-unoccupied” channels.
2. Baseband filters “zoom in” onto those channels and multiple ADCs digitize them.

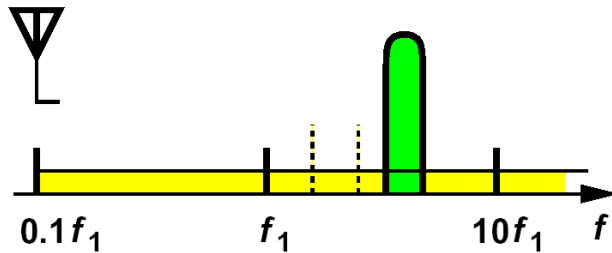


- Given certain blocker levels, what ADC resolution suffices for the first step?

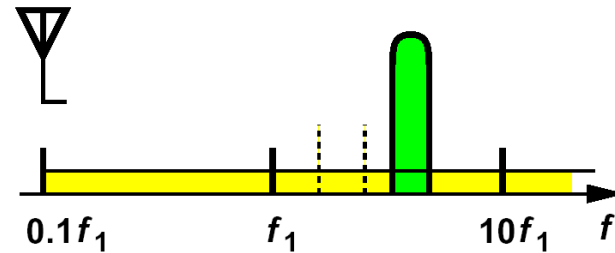
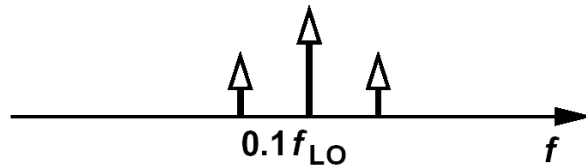
- What criteria should be used to determine “potentially-unoccupied” channels?



# Effect of Spurs, Harmonics, and Other Blemishes



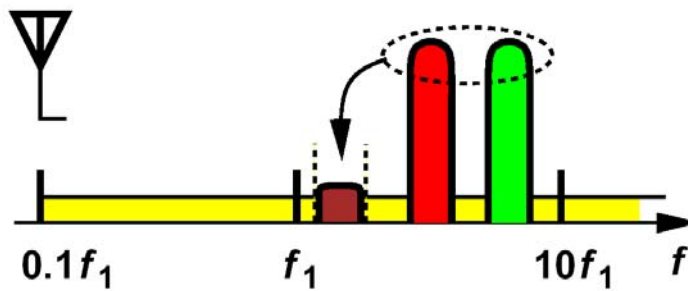
LO Spurs



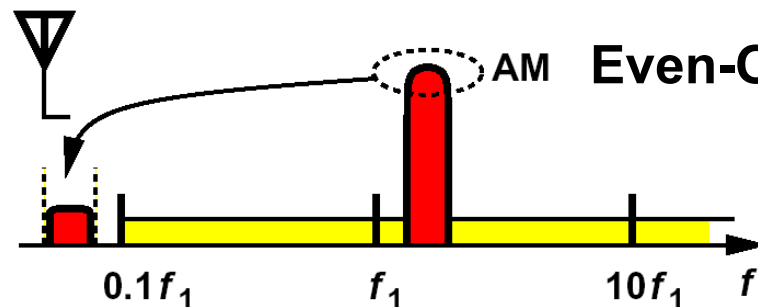
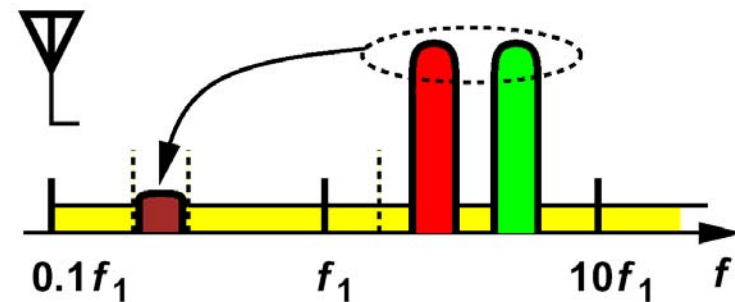
LO Harmonics



Odd-Order Nonlin.



Even-Order Nonlin.

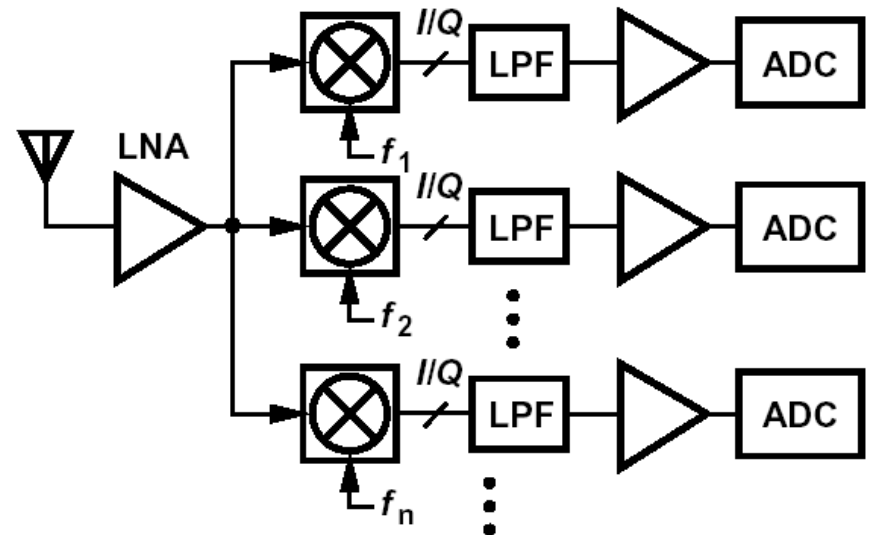


AM Even-Order Nonlin.



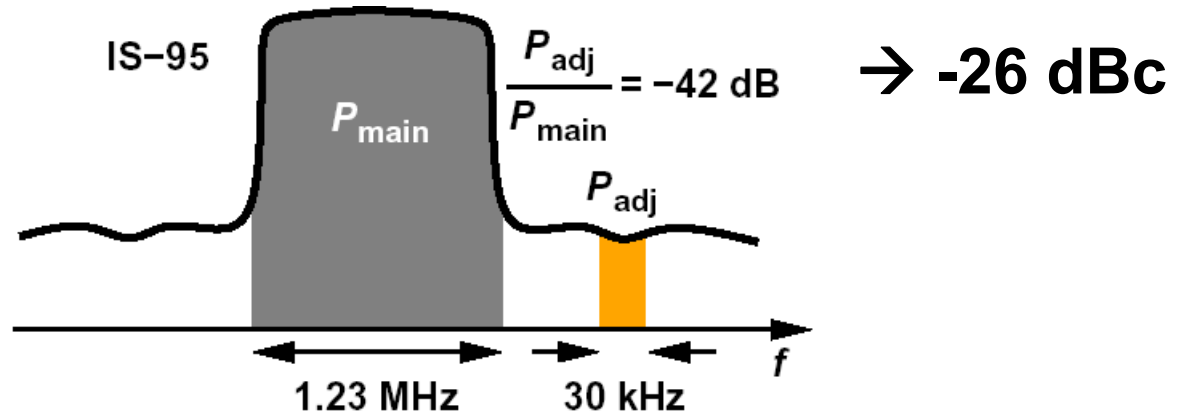
# RX Path

- Broadband gain and input matching
  - Difficult to switch different circuits in and out at the input.
- Low noise – especially flicker noise for 20-50 MHz
- High IP3 and IP2
- Multiple concurrent downconversions to speed up spectrum sensing:



# TX Path

- Broadband upconversion, PA, and matching
- Low adjacent-channel power



- Concurrent Transmission and Sensing?
- Concurrent Transmission in multiple channels?



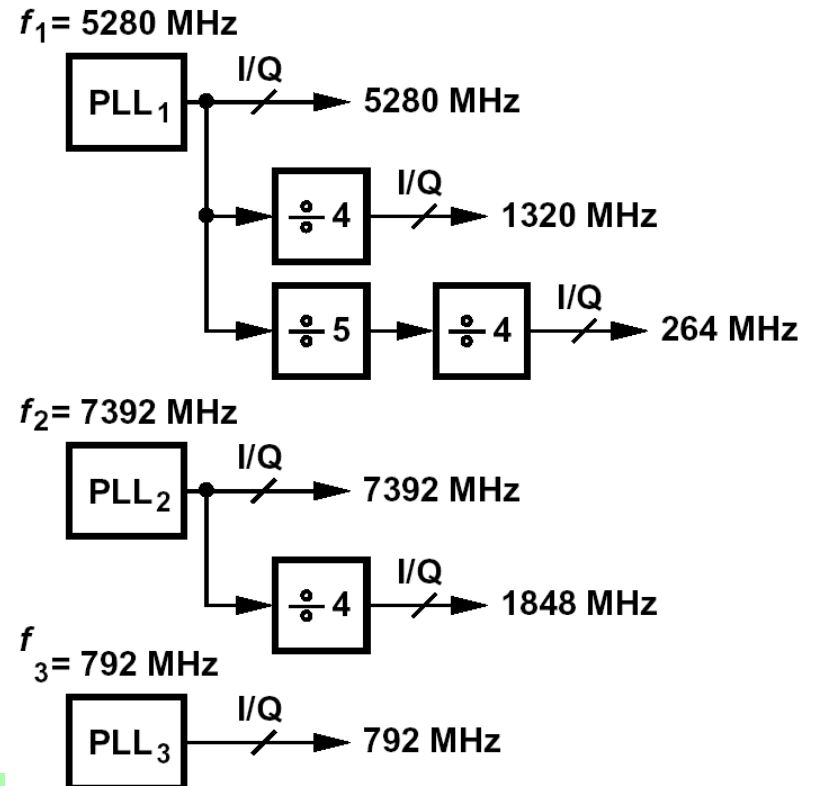
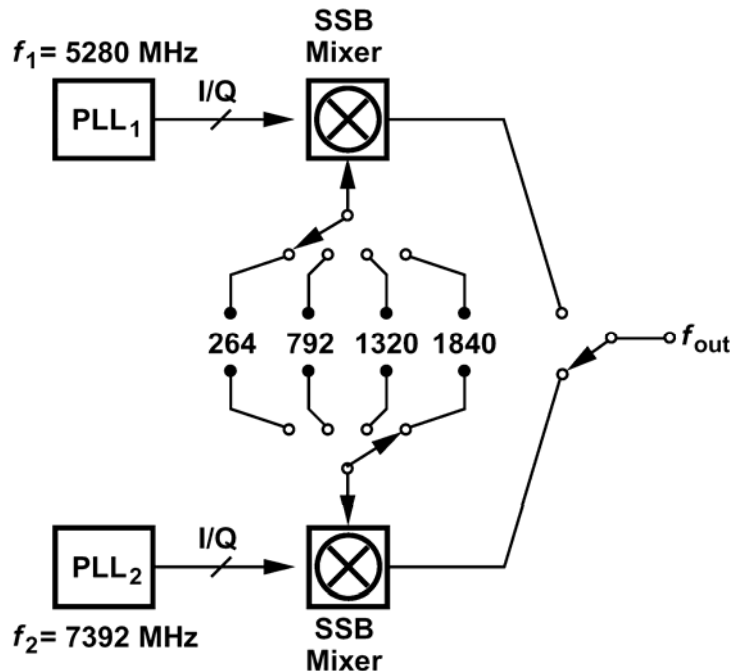
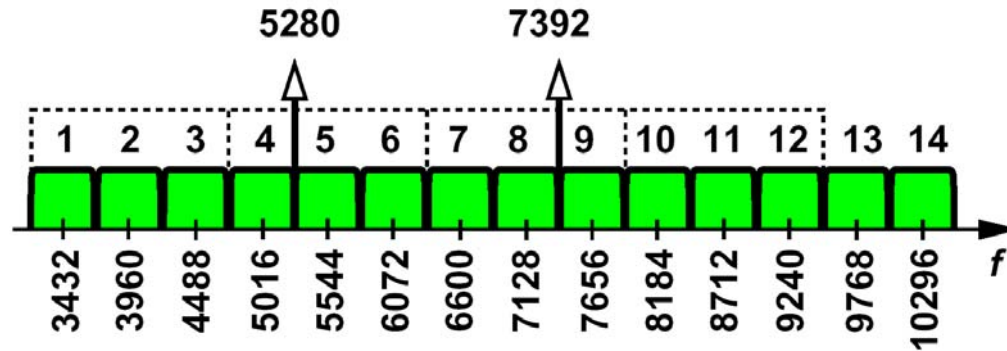
# Frequency Synthesis: 2.5 Decades

---

- LC VCO Tuning Range  $< \pm 10\%$
- If a frequency is divided by an odd number, it must then be divided by 4 to generate quadrature phases.
- Single-sideband mixing probably out of the question
- How many VCOs does it take to cover one decade?



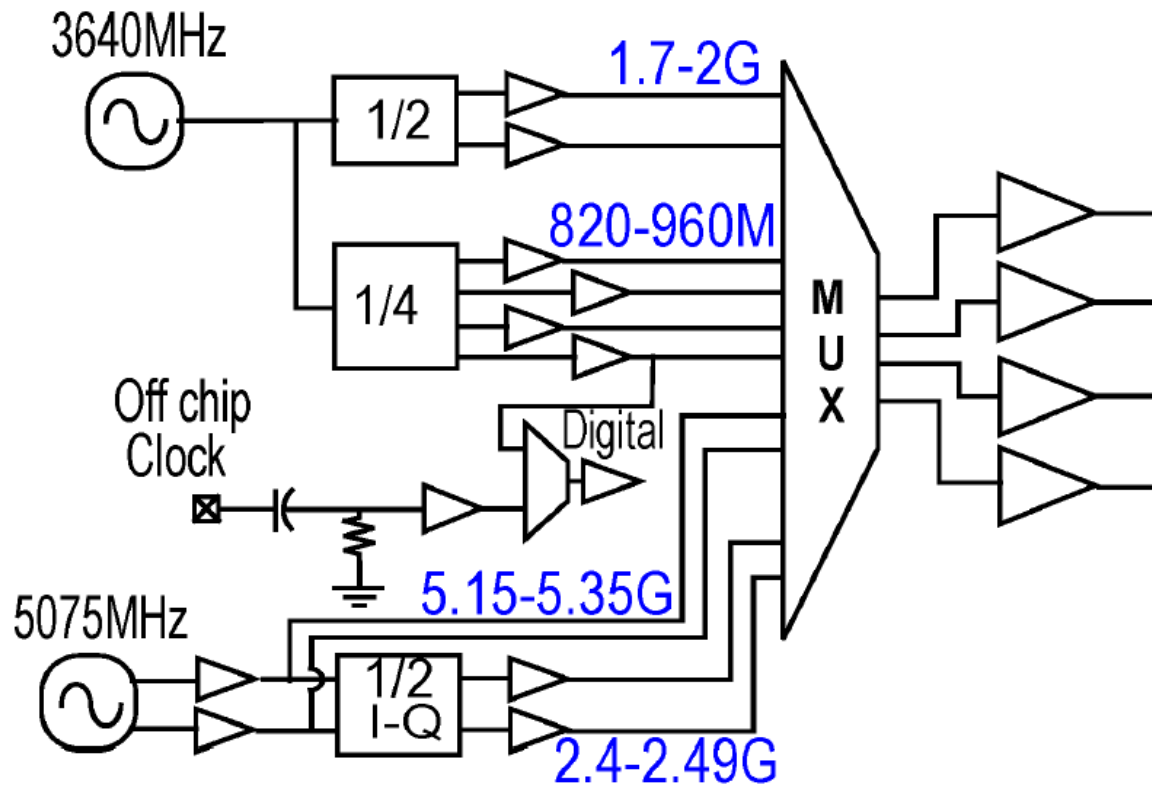
# UWB Example



[Razavi et al, CICC05]



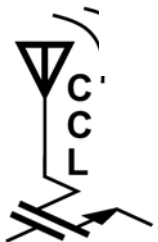
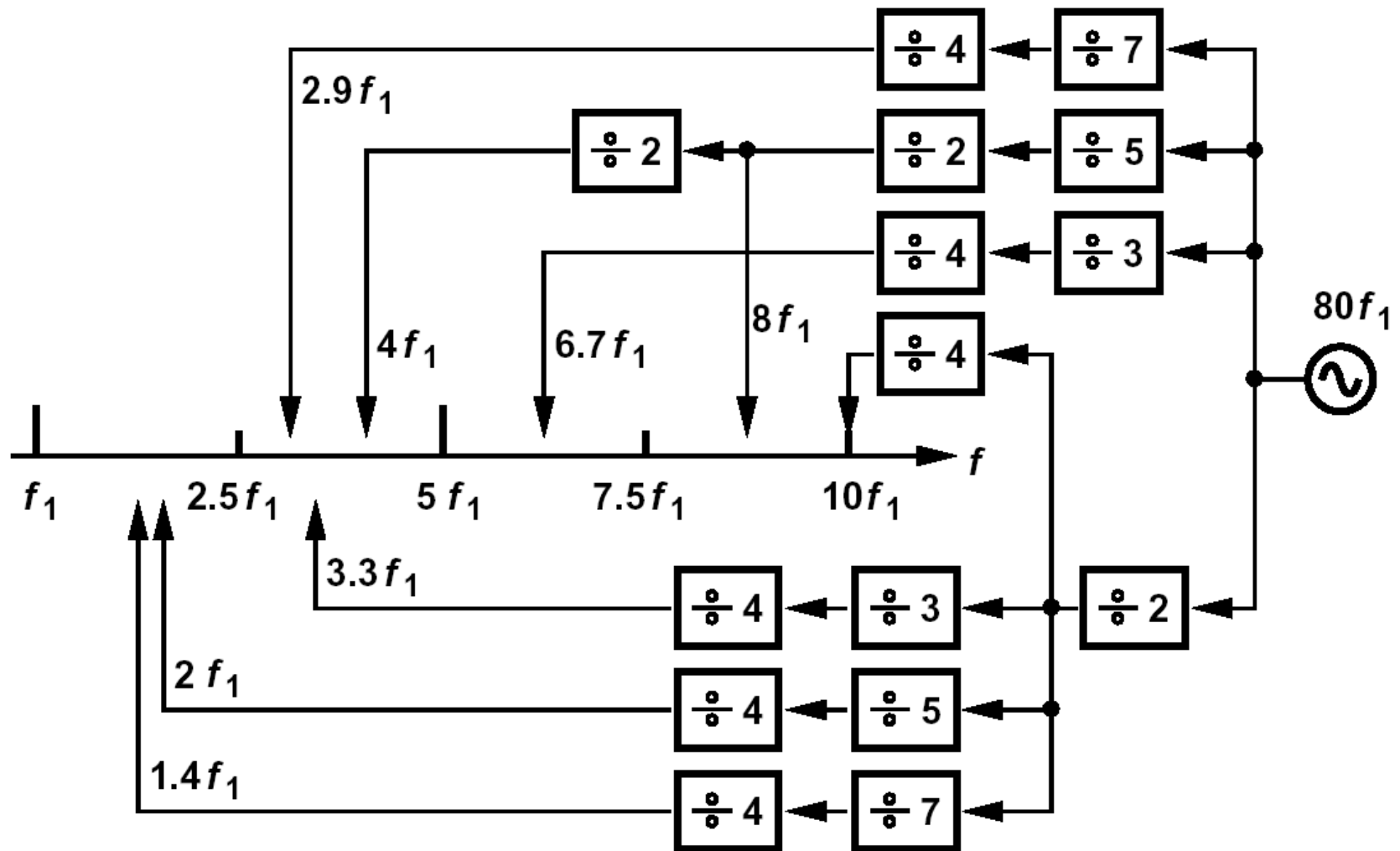
# SDR Example



[Bagheri et al, ISSCC06]

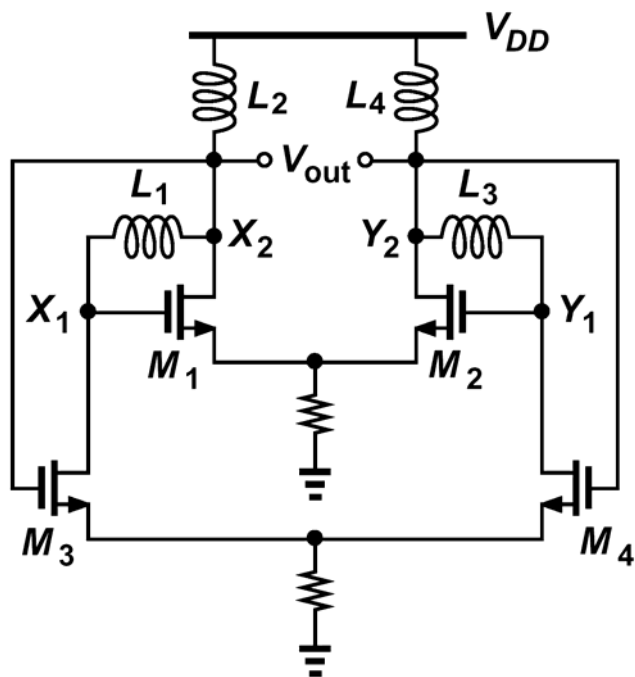
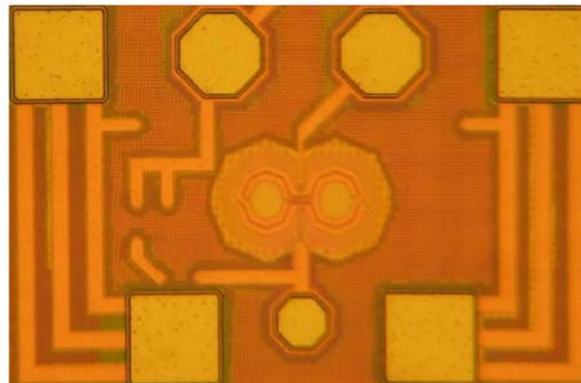


# How to Cover One Decade?

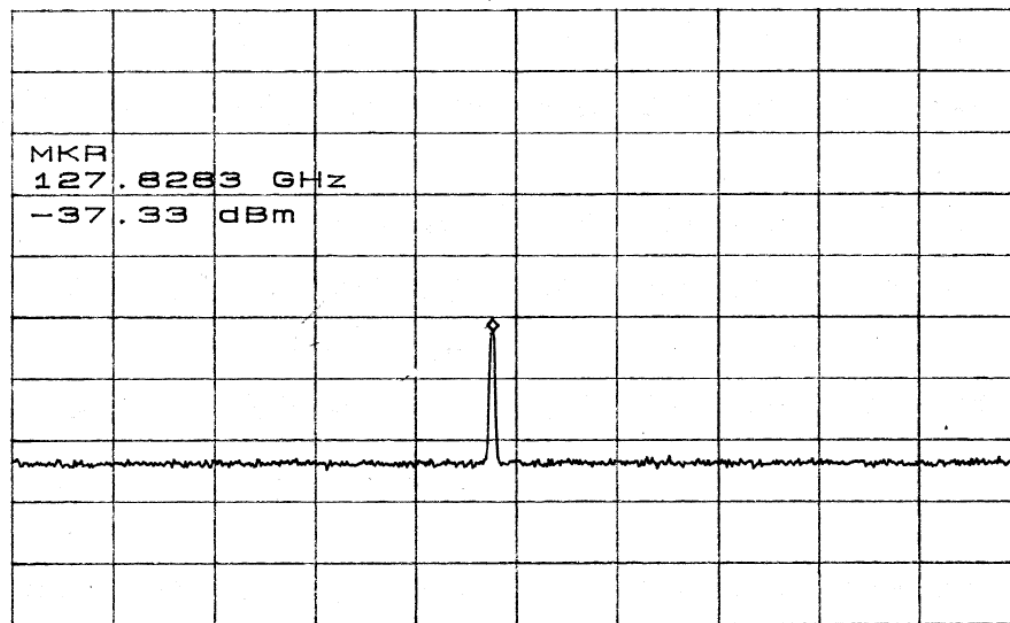


# Millimeter Waves to the Rescue

## 128-GHz Osc. in 90-nm CMOS



CL 45.0dB      MKR -37.33dBm  
RL 15.0dBm      10dB/      127.8283GHz



CENTER 127.8400GHz      SPAN 500.0MHz  
RBW 1.0MHz      \*VBW 10kHz      SWP 130ms

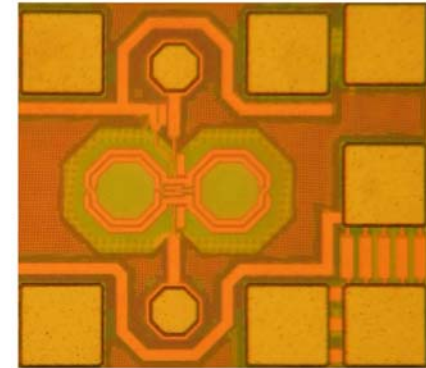
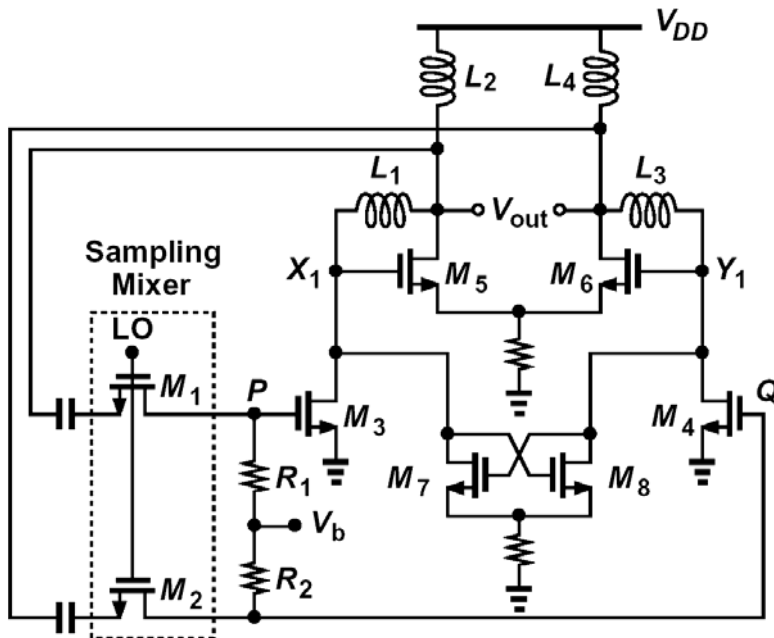
[Razavi, JSSC, Sept. 08]

Communication Circuits Laboratory



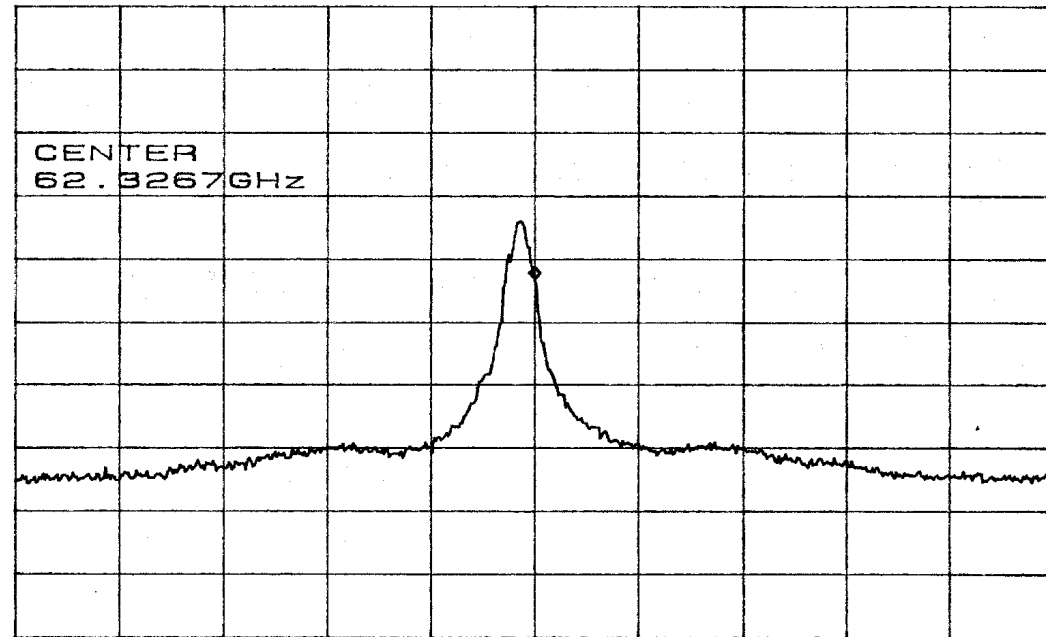
# Millimeter Waves to the Rescue

## 125-GHz Divider in 90-nm CMOS



CL 37.0dB  
RL 7.0dBm

MKR -36.17dBm  
62.3267GHz



CENTER 62.3267GHz

RBW 1.0MHz \*VBW 10kHz

SPAN 100.0MHz

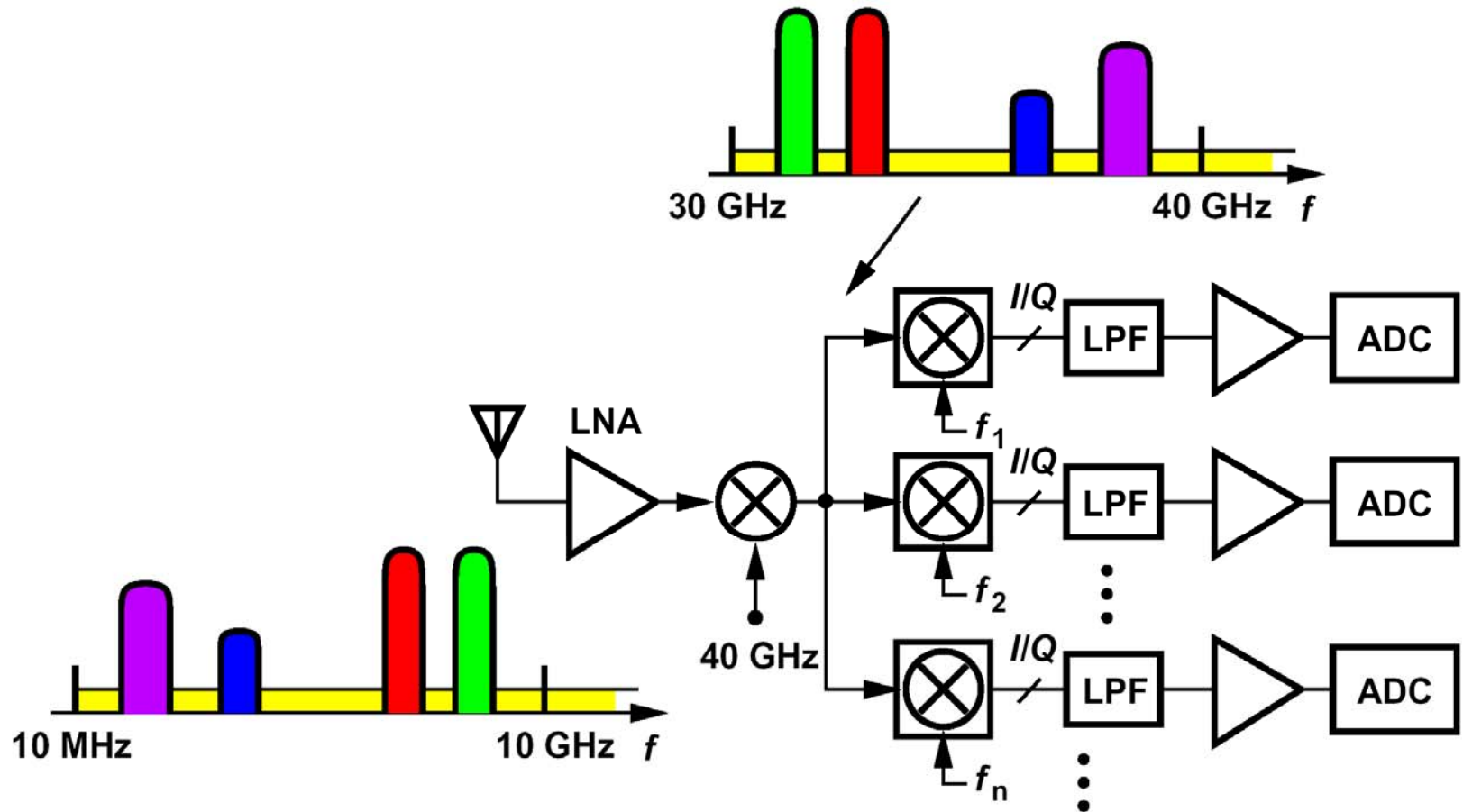
SWP 50.0ms

[Razavi, JSSC, Sept. 08]

Communication Circuits Laboratory



# Alternative Solution



# Conclusion

---

- **Millimeter-wave and cognitive radios pose new challenges in RF and analog design – thereby keeping us employed.**
- **Cross-fertilization of concepts from UWB and mm-waves can greatly benefit CRE design.**
- **Many issues need to be studied and quantified:**
  - **Baseband ADC Requirements**
  - **NF Calibration**
  - **Coverage of 2-3 Frequency Decades**
  - **Broadband Gain, Matching, PAs, etc.**

